

## 250MHz/3mA Current Mode Feedback Amplifier

The EL2480 is a quad current-feedback operational amplifier which achieves a -3dB bandwidth of 250MHz at a gain of +1 while consuming only 3mA of supply current per amplifier. It will operate with dual supplies ranging from  $\pm 1.5\text{V}$  to  $\pm 6\text{V}$ , or from single supplies ranging from +3V to +12V. In spite of its low supply current, the EL2480 can output 55mA while swinging to  $\pm 4\text{V}$  on  $\pm 5\text{V}$  supplies. These attributes make the EL2480 an excellent choice for low power and/or low voltage cable-driver, HDSL, or RGB applications.

For triple applications with disable, consider the EL2386 (16-pin triple).

## Ordering Information

PART NUMBER	PACKAGE	TAPE & REEL	PKG. DWG. #
EL2480CS	14-Pin SO	-	MDP0027
EL2480CS-T7	14-Pin SO	7"	MDP0027
EL2480CS-T13	14-Pin SO	13"	MDP0027
EL2480CSZ (See Note)	14-Pin SO (Pb-free)	-	MDP0027
EL2480CSZ-T7 (See Note)	14-Pin SO (Pb-free)	7"	MDP0027
EL2480CSZ-T13 (See Note)	14-Pin SO (Pb-free)	13"	MDP0027

NOTE: Intersil Pb-free products employ special Pb-free material sets; molding compounds/die attach materials and 100% matte tin plate termination finish, which are RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

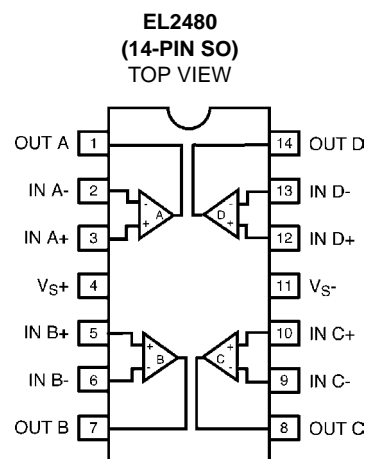
## Features

- Quad topology
- 3mA supply current (per amplifier)
- 250MHz -3dB bandwidth
- Low cost
- Single- and dual-supply operation down to  $\pm 1.5\text{V}$
- 0.05%/0.05° diff. gain/diff. phase into  $150\Omega$
- 1200V/ $\mu\text{s}$  slew rate
- Large output drive current - 55mA
- Also available with disable in triple
- Pb-Free plus Anneal available (RoHS compliant)

## Applications

- Low power/battery applications
- HDSL amplifiers
- Video amplifiers
- Cable drivers
- RGB amplifiers
- Test equipment amplifiers
- Current to voltage converters

## Pinout



**Absolute Maximum Ratings** ( $T_A = 25^\circ\text{C}$ )

Supply Voltage between  $V_{S+}$  and GND. . . . . +12.6V  
 Voltage between  $V_{S+}$  and  $V_{S-}$  . . . . . +12.6V  
 Common-Mode Input Voltage . . . . .  $V_{S-}$  to  $V_{S+}$   
 Differential Input Voltage . . . . .  $\pm 6\text{V}$   
 Current into  $+I_N$  or  $-I_N$  . . . . .  $\pm 7.5\text{mA}$

Internal Power Dissipation . . . . . See Curves  
 Operating Ambient Temperature Range . . . . .  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$   
 Operating Junction Temperature  
 Plastic Packages . . . . .  $150^\circ\text{C}$   
 Output Current . . . . .  $\pm 60\text{mA}$   
 Storage Temperature Range . . . . .  $-65^\circ\text{C}$  to  $+150^\circ\text{C}$

**CAUTION:** Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

**IMPORTANT NOTE:** All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore:  $T_J = T_C = T_A$

**DC Electrical Specifications**  $V_S = \pm 5\text{V}$ ,  $R_L = 150\Omega$ ,  $T_A = 25^\circ\text{C}$  unless otherwise specified

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
$V_{OS}$	Input Offset Voltage			2.5	10	mV
$TCV_{OS}$	Average Input Offset Voltage Drift	Measured from $T_{MIN}$ to $T_{MAX}$		5		$\mu\text{V}/^\circ\text{C}$
$dV_{OS}$	$V_{OS}$ Matching			0.5		mV
$+I_N$	+Input Current			1.5	15	$\mu\text{A}$
$d+I_N$	$+I_N$ Matching			20		nA
$-I_N$	-Input Current			16	40	$\mu\text{A}$
$d-I_N$	$-I_N$ Matching			2		$\mu\text{A}$
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5\text{V}$	45	50		dB
-ICMR	-Input Current Common Mode Rejection	$V_{CM} = \pm 3.5\text{V}$		5	30	$\mu\text{A}/\text{V}$
PSRR	Power Supply Rejection Ratio	$V_S$ is moved from $\pm 4\text{V}$ to $\pm 6\text{V}$	60	70		dB
-IPSR	- Input Current Power Supply Rejection	$V_S$ is moved from $\pm 4\text{V}$ to $\pm 6\text{V}$		1	15	$\mu\text{A}/\text{V}$
$R_{OL}$	Transimpedance	$V_{OUT} = \pm 2.5\text{V}$	120	300		$\text{k}\Omega$
$+R_{IN}$	+Input Resistance	$V_{CM} = \pm 3.5\text{V}$	0.5	2		$\text{M}\Omega$
$+C_{IN}$	+Input Capacitance			1.2		pF
CMIR	Common Mode Input Range		$\pm 3.5$	$\pm 4.0$		V
$V_O$	Output Voltage Swing	$V_S = \pm 5$	$\pm 3.5$	$\pm 4.0$		V
		$V_S = 5$ single-supply, high		4.0		V
		$V_S = 5$ single-supply, low		0.3		V
$I_O$	Output Current	Per amplifier	50	55		mA
$I_S$	Supply Current	Per amplifier		3	6	mA

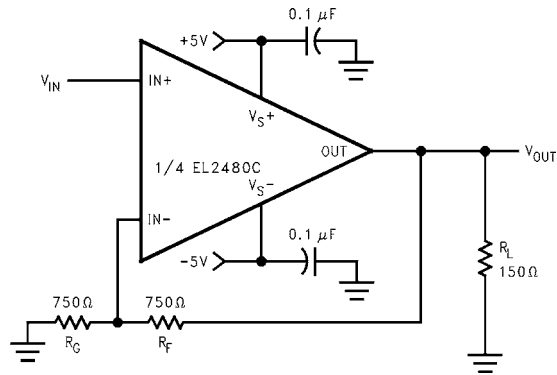
**AC Electrical Specifications**  $V_S = \pm 5V$ ,  $R_F = R_G = 750\Omega$ ,  $R_L = 150\Omega$ ,  $T_A = 25^\circ C$  unless otherwise specified

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
-3dB BW	-3dB Bandwidth	$A_V = 1$		250		MHz
-3dB BW	-3dB Bandwidth	$A_V = 2$		180		MHz
0.1dB BW	0.1dB Bandwidth	$A_V = 2$		50		MHz
SR	Slew Rate	$V_{OUT} = \pm 2.5V$ , $A_V = 2$	600	1200		V/ $\mu s$
$t_R$ , $t_F$	Rise and Fall Time	$V_{OUT} = \pm 500mV$		1.5		ns
$t_{PD}$	Propagation Delay	$V_{OUT} = \pm 500mV$		1.5		ns
OS	Overshoot	$V_{OUT} = \pm 500mV$		3.0		%
$t_S$	0.1% Settling	$V_{OUT} = \pm 2.5V$ , $A_V = -1$		15		ns
dG	Differential Gain	$A_V = 2$ , $R_L = 150\Omega$ (Note 1)		0.05		%
dP	Differential Phase	$A_V = 2$ , $R_L = 150\Omega$ (Note 1)		0.05		°
dG	Differential Gain	$A_V = 1$ , $R_L = 500\Omega$ (Note 1)		0.01		%
dP	Differential Phase	$A_V = 1$ , $R_L = 500\Omega$ (Note 1)		0.01		°
$C_S$	Channel Separation	$f = 5MHz$		85		dB

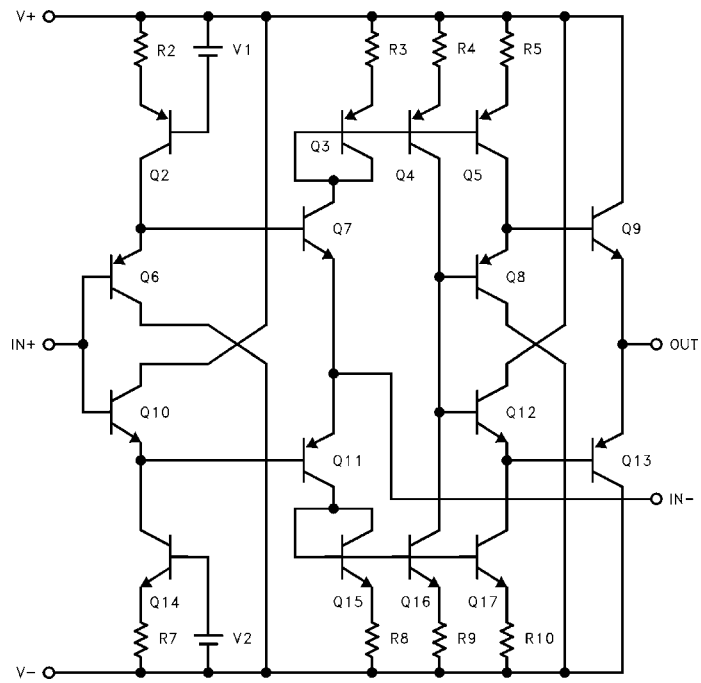
NOTE:

1. DC offset from 0V to 0.714V, AC amplitude 286mV<sub>P-P</sub>,  $f = 3.58MHz$

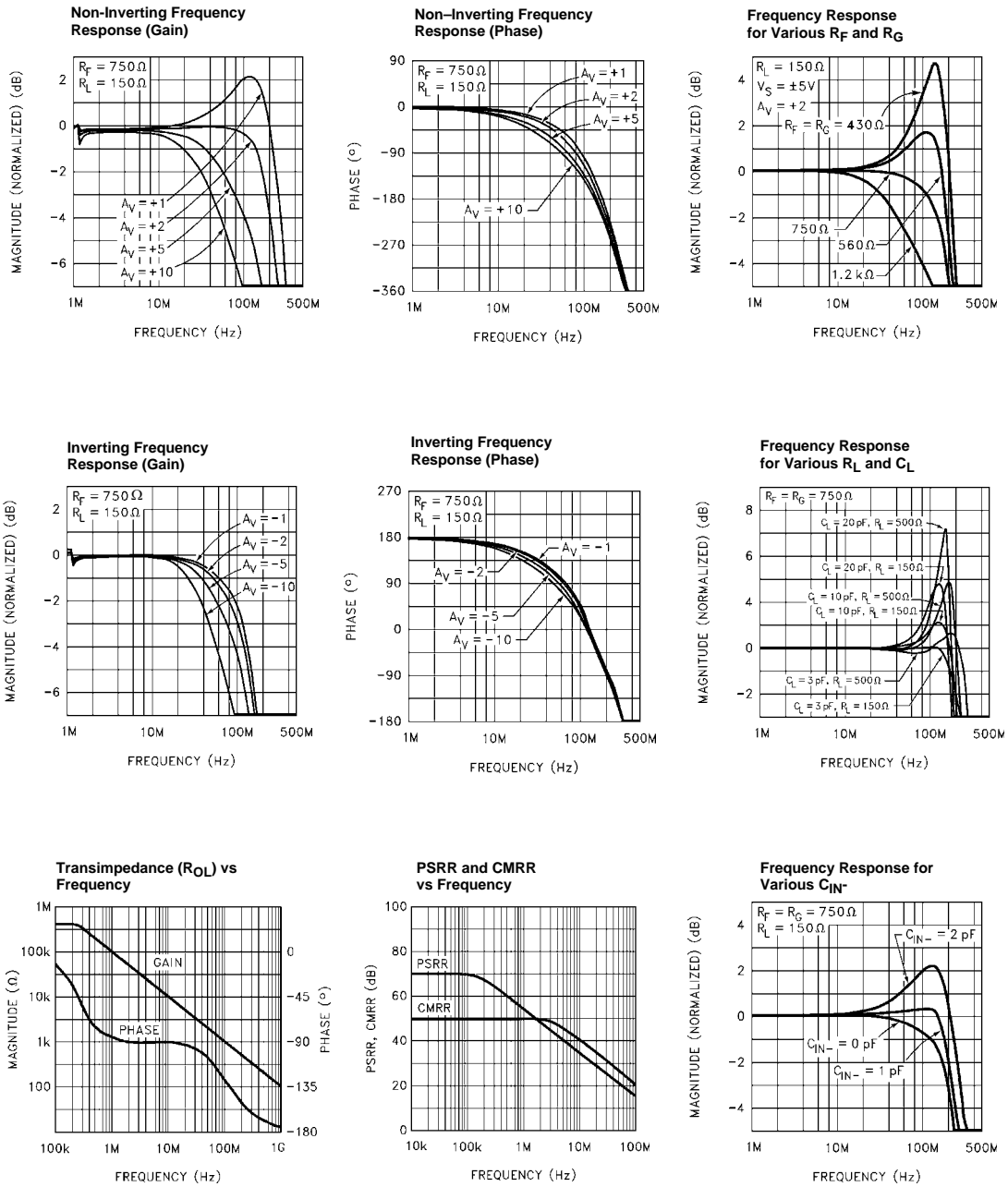
**Test Circuit** (per Amplifier)



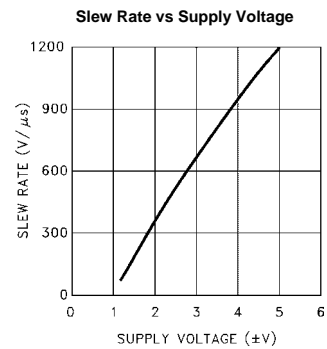
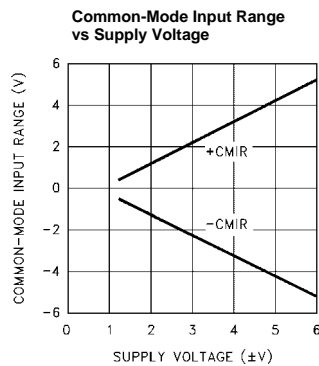
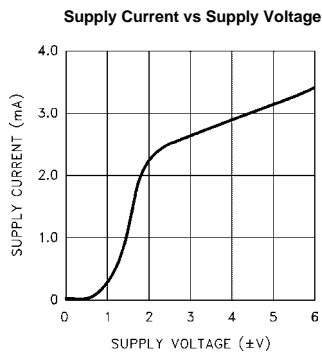
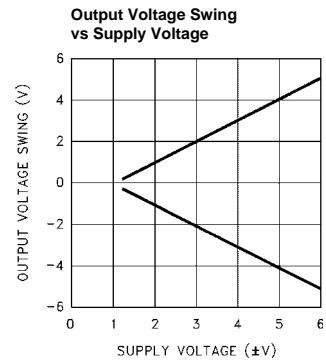
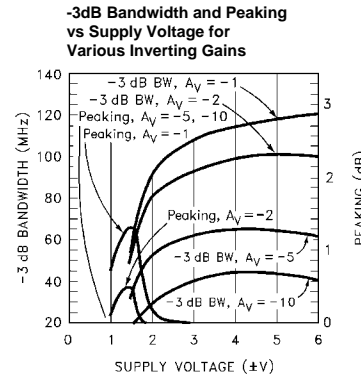
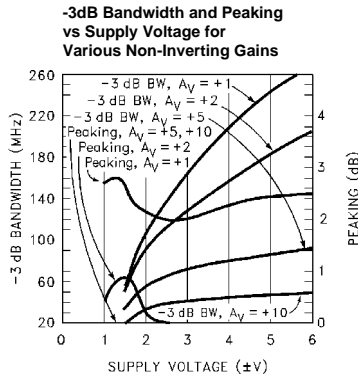
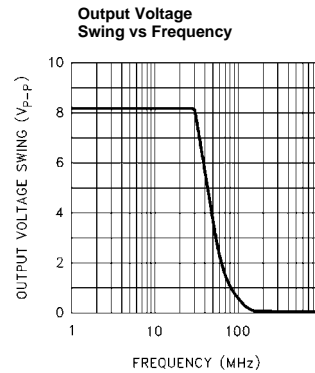
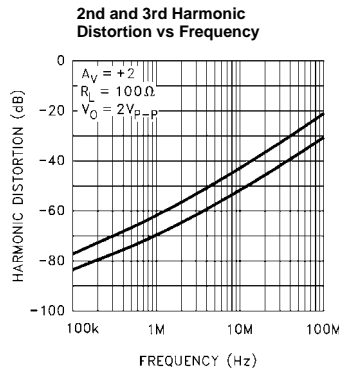
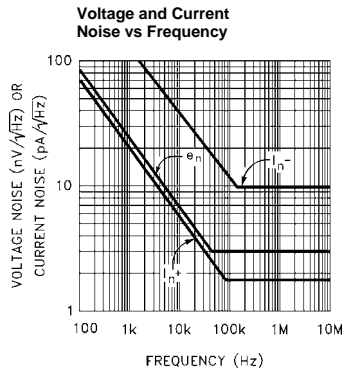
**Simplified Schematic** (per Amplifier)



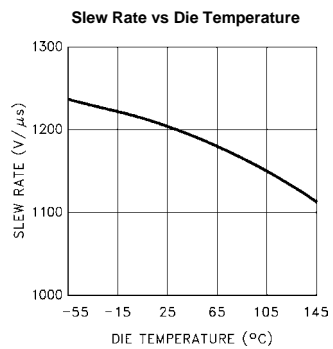
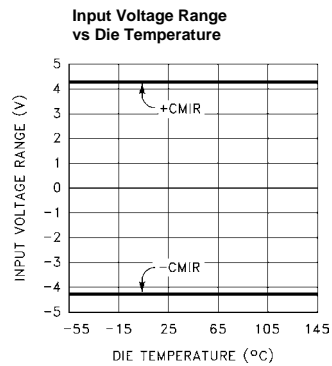
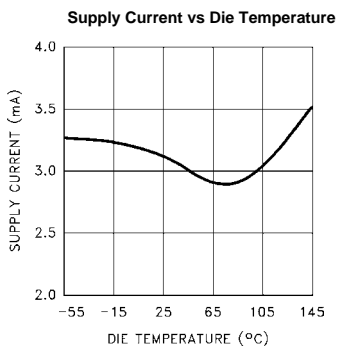
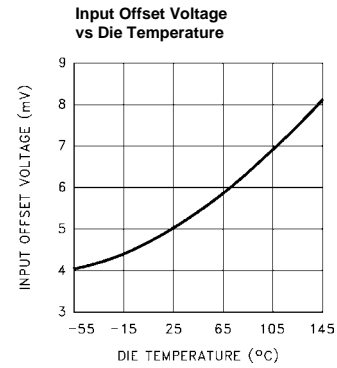
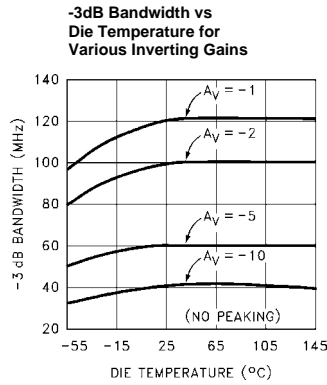
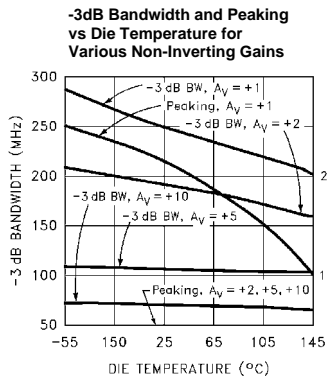
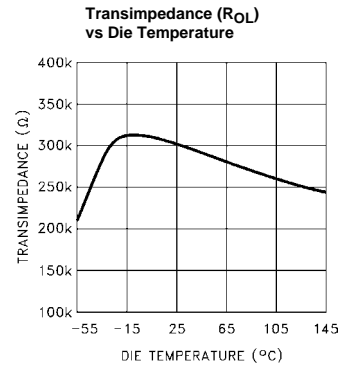
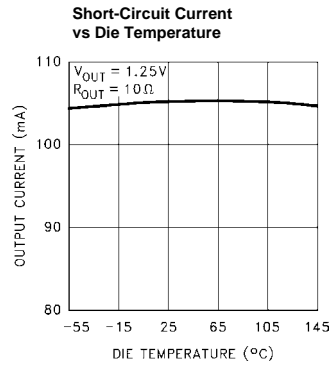
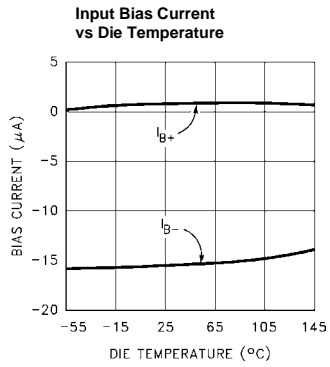
## Typical Performance Curves



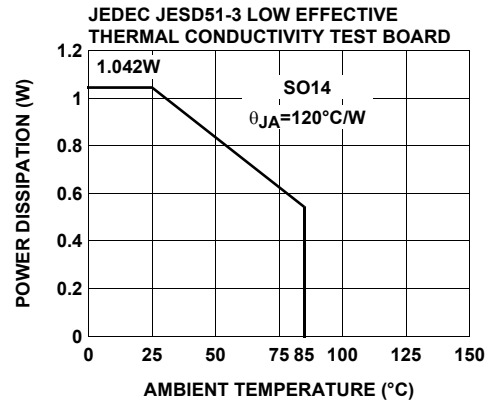
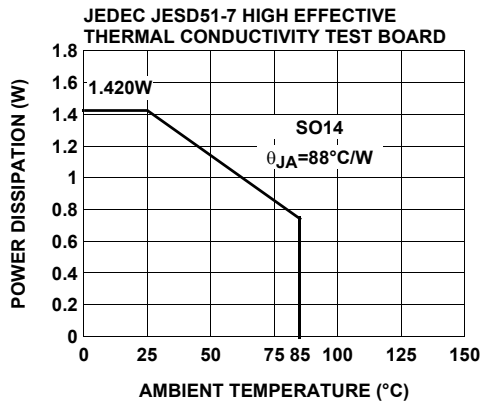
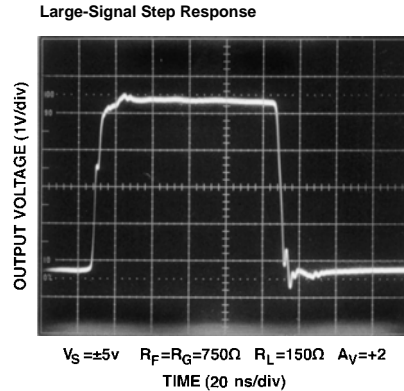
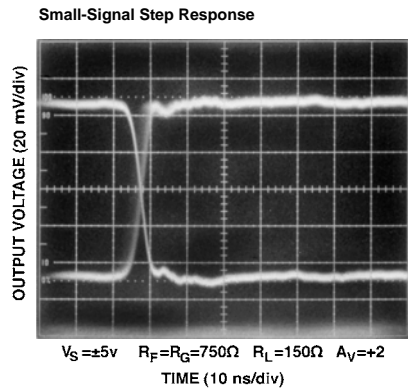
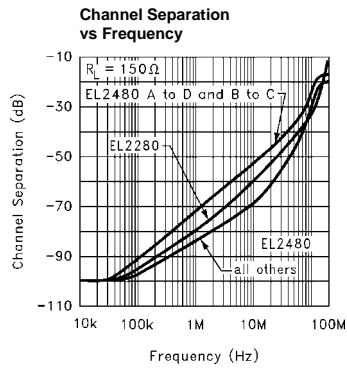
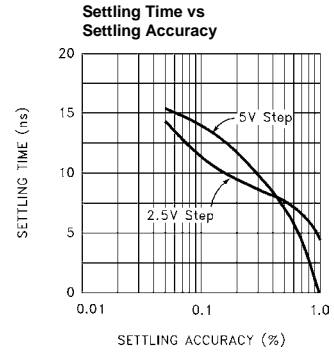
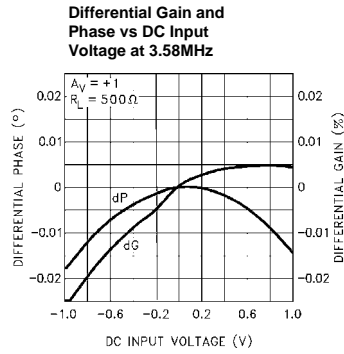
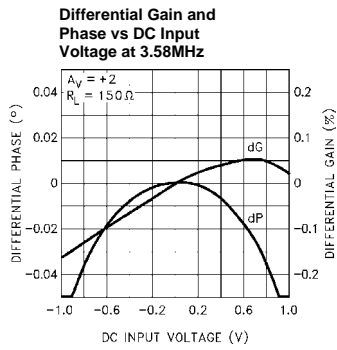
# Typical Performance Curves (Continued)



# Typical Performance Curves (Continued)



# Typical Performance Curves (Continued)





## Applications Information

### Product Description

The EL2480 is a current-feedback operational amplifier that offers a wide -3dB bandwidth of 250MHz and a low supply current of 3mA per amplifier. This product also features high output current drive. The EL2480 can output 55mA per amplifier. The EL2480 works with supply voltages ranging from a single 3V to  $\pm 6V$ , and it is also capable of swinging to within 1V of either supply on the input and the output. Because of its current-feedback topology, the EL2480 does not have the normal gain-bandwidth product associated with voltage-feedback operational amplifiers. This allows its -3dB bandwidth to remain relatively constant as closed-loop gain is increased. This combination of high bandwidth and low power, together with aggressive pricing make the EL2480 the ideal choice for many low-power/high-bandwidth applications such as portable computing, HDSL, and video processing.

The EL2480 is available in the industry standard SO package. For triple application with disable, consider the EL2386 (16-pin triple).

### Power Supply Bypassing and Printed Circuit Board Layout

As with any high-frequency device, good printed circuit board layout is necessary for optimum performance. Ground plane construction is highly recommended. Lead lengths should be as short as possible. The power supply pins must be well bypassed to reduce the risk of oscillation. The combination of a 4.7 $\mu F$  tantalum capacitor in parallel with a 0.1 $\mu F$  capacitor has been shown to work well when placed at each supply pin.

For good AC performance, parasitic capacitance should be kept to a minimum especially at the inverting input (see the Capacitance at the Inverting Input section). Ground plane construction should be used, but it should be removed from the area near the inverting input to minimize any stray capacitance at that node. Carbon or Metal-Film resistors are acceptable with the Metal-Film resistors giving slightly less peaking and bandwidth because of their additional series inductance. Use of sockets, should be avoided if possible. Sockets add parasitic inductance and capacitance which will result in some additional peaking and overshoot.

### Capacitance at the Inverting Input

Any manufacturer's high-speed voltage- or current-feedback amplifier can be affected by stray capacitance at the inverting input. For inverting gains this parasitic capacitance has little effect because the inverting input is a virtual ground, but for non-inverting gains this capacitance (in conjunction with the feedback and gain resistors) creates a pole in the feedback path of the amplifier. This pole, if low enough in frequency, has the same destabilizing effect as a zero in the forward open-loop response. The use of large

value feedback and gain resistors further exacerbates the problem by further lowering the pole frequency.

The experienced user with a large amount of PC board layout experience may find in rare cases that the EL2480 has less bandwidth than expected.

The reduction of feedback resistor values (or the addition of a very small amount of external capacitance at the inverting input, e.g. 0.5pF) will increase bandwidth as desired. Please see the curves for Frequency Response for Various  $R_F$  and  $R_G$ , and Frequency Response for Various  $C_{IN}$ .

### Feedback Resistor Values

The EL2480 has been designed and specified at gains of +1 and +2 with  $R_F = 750\Omega$ . These values of feedback resistors give 250MHz of -3dB bandwidth at  $A_V = +1$  with about 2.5dB of peaking, and 180MHz of -3dB bandwidth at  $A_V = +2$  with about 0.1dB of peaking. Since the EL2480 is current-feedback amplifier, it is also possible to change the value of  $R_F$  to get more bandwidth. As seen in the curve of Frequency Response For Various  $R_F$  and  $R_G$ , bandwidth and peaking can be easily modified by varying the value of the feedback resistor.

Because the EL2480 is current-feedback amplifier, its gain-bandwidth product is not a constant for different closed-loop gains. This feature actually allows the EL2480 to maintain about the same -3dB bandwidth, regardless of closed-loop gain. However, as closed-loop gain is increased, bandwidth decreases slightly while stability increases. Since the loop stability is improving with higher closed-loop gains, it becomes possible to reduce the value of  $R_F$  below the specified 560 $\Omega$  and 750 $\Omega$  and still retain stability, resulting in only a slight loss of bandwidth with increased closed-loop gain.

### Supply Voltage Range and Single-Supply Operation

The EL2480 has been designed to operate with supply voltages having a span of greater than 3V, and less than 12V. In practical terms, this means that the EL2480 will operate on dual supplies ranging from  $\pm 1.5V$  to  $\pm 6V$ . With a single-supply, the EL2480 will operate from +3V to +12V.

As supply voltages continue to decrease, it becomes necessary to provide input and output voltage ranges that can get as close as possible to the supply voltages. The EL2480 has an input voltage range that extends to within 1V of either supply. So, for example, on a single +5V supply, the EL2480 has an input range which spans from 1V to 4V. The output range of the EL2480 is also quite large, extending to within 1V of the supply rail. On a  $\pm 5V$  supply, the output is therefore capable of swinging from -4V to +4V. Single-supply output range is even larger because of the increased negative swing due to the external pull-down resistor to ground. On a single +5V supply, output voltage range is about 0.3V to 4V.

## Video Performance

For good video performance, an amplifier is required to maintain the same output impedance and the same frequency response as DC levels are changed at the output. This is especially difficult when driving a standard video load of 150Ω, because of the change in output current with DC level. Until the EL2480, good Differential Gain could only be achieved by running high idle currents through the output transistors (to reduce variations in output impedance). These currents were typically comparable to the entire 3mA supply current of EL2480 amplifier! Special circuitry has been incorporated in the EL2480 to reduce the variation of output impedance with current output. This results in dG and dP specifications of 0.05% and 0.05° while driving 150Ω at a gain of +2.

Video performance has also been measured with a 500Ω load at a gain of +1. Under these conditions, the EL2480 has dG and dP specifications of 0.01% and 0.01° respectively while driving 500Ω at  $A_V = +1$ .

## Output Drive Capability

This amplifier of the EL2480 is capable of providing a minimum of ±50mA. These output drive levels are unprecedented in amplifiers running at these supply currents. The ±50mA minimum output drive of the EL2480 amplifier allows swings of ±2.5V into 50Ω loads.

## Driving Cables and Capacitive Loads

When used as a cable driver, double termination is always recommended for reflection-free performance. For those applications, the back-termination series resistor will decouple the EL2480 from the cable and allow extensive capacitive drive. However, other applications may have high capacitive loads without a back-termination resistor. In these applications, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output to eliminate most peaking. The gain resistor ( $R_G$ ) can then be chosen to make up for any gain loss which may be created by this additional resistor at the output. In many cases it is also possible to simply increase the value of the feedback resistor ( $R_F$ ) to reduce the peaking.

## Current Limiting

The EL2480 has no internal current-limiting circuitry. If any output is shorted, it is possible to exceed the Absolute Maximum Ratings for output current or power dissipation, potentially resulting in the destruction of the device.

## Power Dissipation

With the high output drive capability of the EL2480, it is possible to exceed the 150°C Absolute Maximum junction temperature under certain very high load current conditions. Generally speaking, when  $R_L$  falls below about 25Ω, it is important to calculate the maximum junction temperature ( $T_{JMAX}$ ) for the application to determine if power-supply voltages, load conditions, or package type need to be

modified for the EL2480 to remain in the safe operating area. These parameters are calculated as follows:

$$T_{JMAX} = T_{MAX} + (\theta_{JA} \times n \times PD_{MAX})$$

where:

$T_{MAX}$  = Maximum ambient temperature

$\theta_{JA}$  = Thermal resistance of the package

$n$  = Number of amplifiers in the package

$PD_{MAX}$  = Maximum power dissipation of each amplifier in the package

$PD_{MAX}$  for each amplifier can be calculated as follows:

$$PD_{MAX} = (2 \times V_S \times I_{SMAX}) + (V_S - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_L}$$

where:

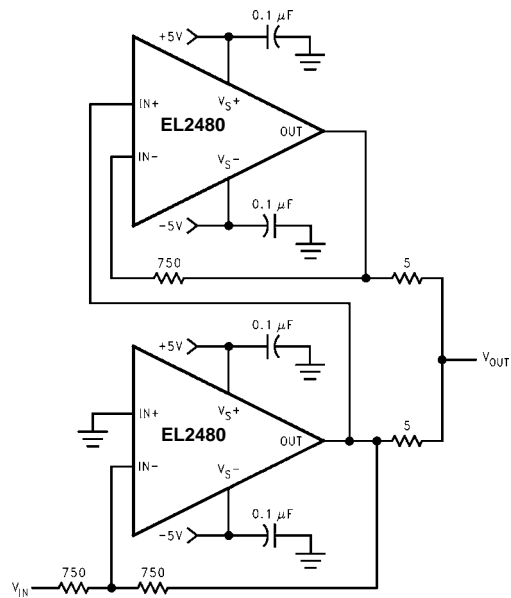
$V_S$  = Supply voltage

$I_{SMAX}$  = Maximum supply current of 1 amplifier

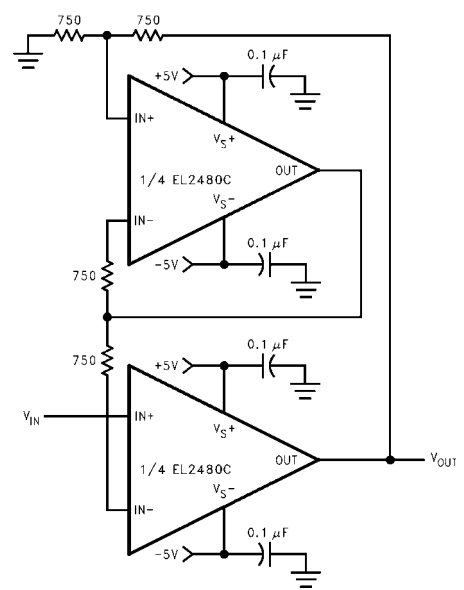
$V_{OUTMAX}$  = Maximum output voltage of the application

$R_L$  = Load resistance

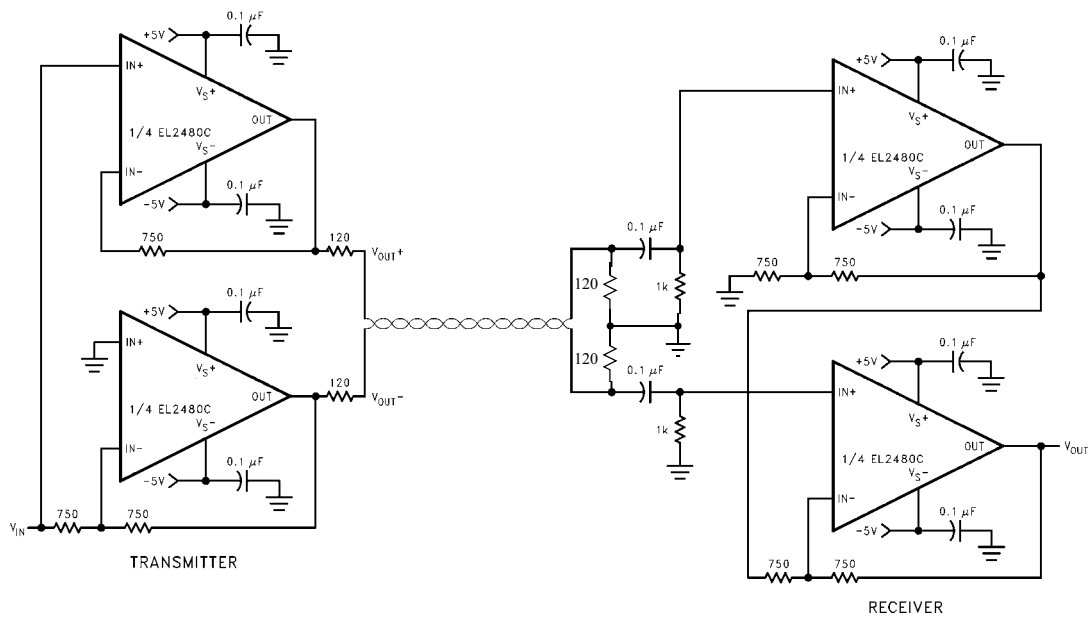
## Typical Application Circuits



**INVERTING 200mA OUTPUT CURRENT DISTRIBUTION AMPLIFIER**



**FAST-SETTLING PRECISION AMPLIFIER**



**DIFFERENTIAL LINE-DRIVER/RECEIVER**

## EL2480 Macromodel

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* EL2480 Macromodel
* Revision A, March 1995
* AC characteristics used: Rf = Rg = 750Ω
* Connections:
*
*      +input
*      |
*      | -input
*      |
*      | +Vsupply
*      |
*      | -Vsupply
*      |
*      | output
*
.subckt EL2480/el 3 2 4 11 1
*
* Input Stage
*
e1 10 0 3 0 1.0
vis 10 9 0V
h2 9 12 vxx 1.0
r1 2 11 400
l1 11 12 25nH
iinp 3 0 1.5uA
iinm 2 0 3uA
r12 3 0 2Meg
*
* Slew Rate Limiting
*
h1 13 0 vis 600
r2 13 14 1K
d1 14 0 dclamp
d2 0 14 dclamp
*
* High Frequency Pole
*
e2 30 0 14 0 0.001666666666
l3 30 17 150nH
c5 17 0 0.8pF
r5 17 0 165
*
* Transimpedance Stage
*
g1 0 18 17 0 1.0
rol 18 0 450K
cdp 18 0 0.675pF
*
* Output Stage
*
q1 11 18 19 qp
q2 4 18 20 qn
q3 4 19 21 qn
q4 11 20 22 qp
r7 21 1 4
r8 22 1 4
ios1 4 19 1mA
ios2 20 11 1mA
*
* Supply Current
*
ips 4 11 0.2mA
*
* Error Terms

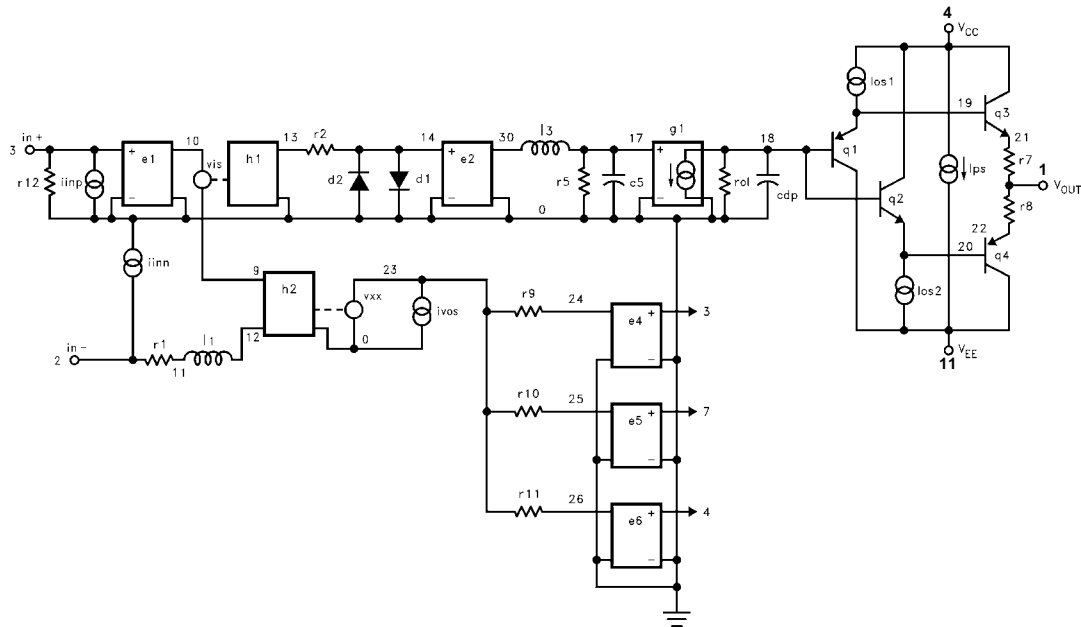
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*
ivos 0 23 0.2mA
vxx 23 0 0V
e4 24 0 3 0 1.0
e5 25 0 4 0 1.0
e6 26 0 11 0 -1.0
r9 24 23 316
r10 25 23 3.2K
r11 26 23 3.2K
*
* Models
*
.model qn npn(is=5e-15 bf=200 tf=0.01nS)
*.model qp pnp(is=5e-15 bf=200 tf=0.01nS)
.model dclamp d(is=1e-30 ibv=0.266
+ bv=0.71v n=4)
.ends

```

## EL2480 Macromodel



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