

## General Description

The LP2950 and LP2951 are micropower voltage regulators with very low dropout voltage (typically 40mV at light loads and 380mV at 100mA), and very low quiescent current (75µA typical). The quiescent current of the LP2950/LP2951 increases only slightly in dropout, thus prolonging battery life. This feature, among others, makes the LP2950 and LP2951 ideally suited for use in battery-powered systems.

Available in a 3-Pin TO-92 package, the LP2950 is pin-compatible with the older 5V regulators. Additional system functions, such as programmable output voltage and logic-controlled shutdown, are available in the 8-pin DIP and 8-pin SOIC versions of the LP2951.

## Applications

- Automotive Electronics
- Voltage Reference
- Avionics

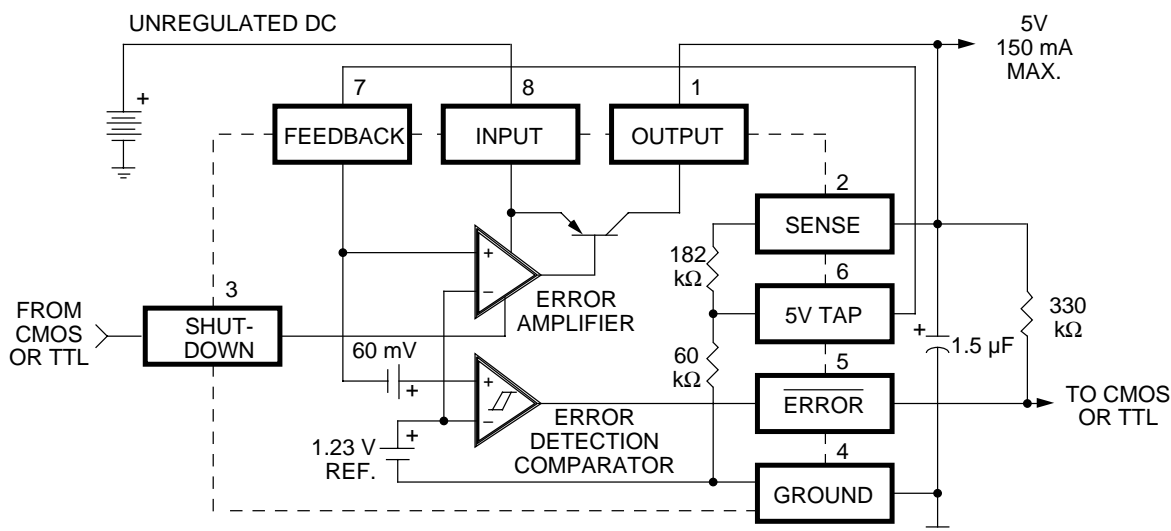
## Features

- High accuracy 5V, guaranteed 100 mA output
- Extremely low quiescent current
- Low-dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Use as regulator or reference
- Needs only 1µF for stability
- Current and thermal limiting

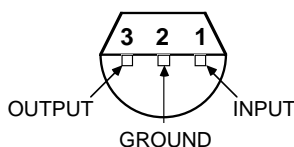
## LP2951 Versions Only

- Error flag warns of output dropout
- Logic-controlled electronic shutdown
- Output programmable from 1.24 to 29V

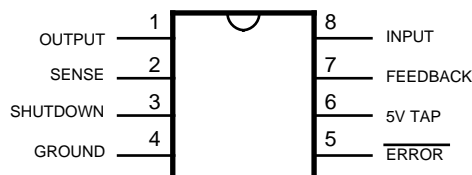
## Block Diagram and Pin Configurations



LP2950 and LP2951 Block Diagram  
(Pin Numbers Refer to LP2951)



TO-92 Plastic Package Bottom View  
(BZ)



DIP and SO Packages  
(BN and BM)

See MIC2950 for a part with 1) higher output (150 mA), 2) transient protection (60V), and 3) reverse input protection to -20V

Additional features available with the LP2951 also include an error flag output that warns of a low output voltage, which is often due to failing batteries on the input. This may also be used as a power-on reset. A logic-compatible shutdown input is also available which enables the regulator to be switched on and off. This part may also be pin-strapped for a 5V output, or programmed from 1.24V to 29V with the use of two external resistors.

The LP2950 is available as either an -02 or -03 version. The -02 and -03 versions are guaranteed for junction temperatures

from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ; the -02 version has a tighter output and reference voltage specification range over temperature. The LP2951 is available as an -02 or -03 version.

The LP2950 and LP2951 have a tight initial tolerance (0.5% typical), a very low output voltage temperature coefficient which allows use as a low-power voltage reference, and extremely good load and line regulation (0.05% typical). This greatly reduces the error in the overall circuit, and is the result of careful design techniques and process control.

## Ordering Information

Part Number	Voltage	Temperature Range*	Package	Accuracy
LP2950-02BZ	5.0V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	3-Pin TO-92 plastic	0.5%
LP2950-03BZ	5.0V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	3-Pin TO-92 plastic	1.0%
LP2951-02BM	5.0V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-Pin SOIC	0.5%
LP2951-03BM	5.0V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-Pin SOIC	1.0%
LP2951-02BN	5.0V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-Pin Plastic DIP	0.5%
LP2951-03BN	5.0V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-Pin Plastic DIP	1.0%
LP2951-4.8BM	4.85V	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-Pin SOIC	1.0%

\* Junction temperatures

## Absolute Maximum Ratings

**If Military/Aerospace specified devices are required, contact your local Micrel representative/distributor for availability and specifications.**

Power dissipation	Internally Limited
Lead Temperature (Soldering, 5 seconds)	$260^{\circ}\text{C}$
Storage Temperature Range	$-65^{\circ}\text{C}$ to $+150^{\circ}\text{C}$
Operating Junction Temperature Range (Note 8)	
LP2950, LP2951	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$
Input Supply Voltage	$-0.3\text{V}$ to $+30\text{V}$
Feedback Input Voltage (Notes 9 and 10)	$-1.5\text{V}$ to $+30\text{V}$
Shutdown Input Voltage (Note 9)	$-0.3\text{V}$ to $+30\text{V}$
Error Comparator Output Voltage (Note 9)	$-0.3\text{V}$ to $+30\text{V}$
ESD Rating is to be determined.	

**Electrical Characteristics** Note 1  $T_A = 25^\circ\text{C}$  except as noted.

Parameter	Condition	Min	Typ	Max	Units
Output Voltage $T_J = 25^\circ\text{C}$	LP295x-02 ( $\pm 0.5\%$ )	4.975	5.000	5.025	V
	LP295x-03 ( $\pm 1\%$ )	4.950	5.000	5.050	V
	LP2951-4.8 ( $\pm 1\%$ )	4.802	4.850	4.899	V
Output Voltage $-25^\circ\text{C} \leq T_J \leq +85^\circ\text{C}$	LP295x-02 ( $\pm 0.5\%$ )	4.950		5.050	V
	LP295x-03 ( $\pm 1\%$ )	4.925		5.075	V
	LP2951-4.8 ( $\pm 1\%$ )	4.777		4.872	V
Output Voltage Over Full Temperature Range	LP295x-02 ( $\pm 0.5\%$ ), $-40^\circ\text{C}$ to $+125^\circ\text{C}$	<b>4.940</b>		<b>5.060</b>	V
	LP295x-03 ( $\pm 1\%$ ), $-40^\circ\text{C}$ to $+125^\circ\text{C}$	<b>4.900</b>		<b>5.100</b>	V
	LP2951-4.8 ( $\pm 1\%$ ), $-40^\circ\text{C}$ to $+125^\circ\text{C}$	<b>4.753</b>		<b>4.947</b>	V
Output Voltage Over Load Variation	LP295x-02 ( $\pm 0.5\%$ ), $100\mu\text{A} \leq I_L \leq 100\text{mA}$ , $T_J \leq T_{J(\text{max})}$	<b>4.930</b>		<b>5.070</b>	V
	LP295x-03 ( $\pm 1\%$ ), $100\mu\text{A} \leq I_L \leq 100\text{mA}$ , $T_J \leq T_{J(\text{max})}$	<b>4.880</b>		<b>5.120</b>	V
	LP2951-4.8 ( $\pm 1\%$ ), $100\mu\text{A} \leq I_L \leq 100\text{mA}$ , $T_J \leq T_{J(\text{max})}$	<b>4.733</b>		<b>4.967</b>	V
Output Voltage Temperature Coefficient	LP295x-02 ( $\pm 0.5\%$ ), Note 12		<b>20</b>	<b>100</b>	ppm/ $^\circ\text{C}$
	LP295x-03 ( $\pm 1\%$ ), Note 12		<b>50</b>	<b>150</b>	ppm/ $^\circ\text{C}$
	LP2951-4.8 ( $\pm 1\%$ ), Note 12		<b>50</b>	<b>150</b>	ppm/ $^\circ\text{C}$
Line Regulation	LP295x-02 ( $\pm 0.5\%$ ), Notes 14, 15		0.03	0.10 <b>0.20</b>	% %
	LP295x-03 ( $\pm 1\%$ ), Notes 14, 15		0.04	0.20 <b>0.40</b>	% %
	LP2951-4.8 ( $\pm 1\%$ ), Notes 14, 15		0.04	0.20 <b>0.40</b>	% %
Load Regulation	LP295x-02 ( $\pm 0.5\%$ ), Note 14, $100\mu\text{A} \leq I_L \leq 100\text{mA}$		0.04	0.10 <b>0.20</b>	% %
	LP295x-03 ( $\pm 1\%$ ), Note 14, $100\mu\text{A} \leq I_L \leq 100\text{mA}$		0.10	0.20 <b>0.30</b>	% %
	LP2951-4.8 ( $\pm 1\%$ ), Note 14, $100\mu\text{A} \leq I_L \leq 100\text{mA}$		0.10	0.20 <b>0.30</b>	% %
Dropout Voltage	Note 5, $I_L = 100\mu\text{A}$		50	80 <b>150</b>	mV mV
	Note 5, $I_L = 100\text{mA}$		380	450 <b>600</b>	mV mV
Ground Current	$I_L = 100\mu\text{A}$		100	150 <b>200</b>	$\mu\text{A}$ $\mu\text{A}$
	$I_L = 100\text{mA}$		8	12 <b>14</b>	mA mA
Dropout Current	$V_{IN} = 4.5\text{V}$ , $I_L = 100\mu\text{A}$		180	250 <b>310</b>	$\mu\text{A}$ $\mu\text{A}$

Parameter	Condition	Min	Typ	Max	Units
Current Limit	$V_{OUT} = 0V$		160	200 <b>220</b>	mA mA
Thermal Regulation	Note 13		0.05	0.20	%/W
Output Noise	10Hz to 100kHz, $C_L = 1\mu F$		430		$\mu V_{RMS}$
	10Hz to 100kHz, $C_L = 200\mu F$		160		$\mu V_{RMS}$
	10Hz to 100kHz, $C_L = 3.3\mu F$ , 0.01 $\mu F$ bypass Feedback to Output		100		$\mu V_{RMS}$
Reference Voltage	LP295x-02 ( $\pm 0.5\%$ )	1.220 <b>1.200</b>	1.235	1.250 <b>1.260</b>	V V
	LP295x-03 ( $\pm 1\%$ )	1.210 <b>1.200</b>	1.235	1.260 <b>1.270</b>	V V
	LP2951-4.8 ( $\pm 1\%$ )	1.210 <b>1.200</b>	1.235	1.260 <b>1.270</b>	V V
Reference Voltage	LP295x-02 ( $\pm 0.5\%$ ), Note 7	<b>1.190</b>		<b>1.270</b>	V
	LP295x-03 ( $\pm 1\%$ ), Note 7	<b>1.185</b>		<b>1.285</b>	V
	LP2951-4.8 ( $\pm 1\%$ ), Note 7	<b>1.185</b>		<b>1.285</b>	V
Feedback Bias Current			20	40 <b>60</b>	nA nA
Reference Voltage	LP295x-02 ( $\pm 0.5\%$ ), Note 12		20		ppm/ $^{\circ}C$
	LP295x-03 ( $\pm 1\%$ ), Note 12		50		ppm/ $^{\circ}C$
	LP2951-4.8 ( $\pm 1\%$ ), Note 12		50		ppm/ $^{\circ}C$
Feedback Bias Current Temperature Coefficient			0.1		nA/ $^{\circ}C$
Output Leakage Current	$V_{OH} = 30V$		0.01	1.00 <b>2.00</b>	$\mu A$ $\mu A$
Output Low Voltage (Flag)	$V_{IN} = 4.5V$ , $I_{OL} = 200\mu A$		150	250 <b>400</b>	mV mV
Upper Threshold Voltage	Note 6	40 <b>25</b>	60		mV mV
Lower Threshold Voltage	Note 6		75	95 <b>140</b>	mV mV
Hysteresis	Note 6		15		mV
Input Logic Voltage	LP295x-02 ( $\pm 0.5\%$ )		1.3		V
	Low			<b>0.7</b>	V
	High	<b>2.0</b>			V
	LP295x-03 ( $\pm 1\%$ )		1.3		V
	Low			<b>0.7</b>	V
	High	<b>2.0</b>			V
	LP2951-4.8 ( $\pm 1\%$ )		1.3		V
	Low			<b>0.7</b>	V
	High	<b>2.0</b>			V

Parameter	Condition	Min	Typ	Max	Units
Shutdown Input Current	$V_{\text{SHUTDOWN}} = 2.4\text{V}$		30	50 <b>100</b>	$\mu\text{A}$ $\mu\text{A}$
	$V_{\text{SHUTDOWN}} = 30\text{V}$		450	600 <b>750</b>	$\mu\text{A}$ $\mu\text{A}$
Regulator Output Current in Shutdown	Note 11		3	10 <b>20</b>	$\mu\text{A}$ $\mu\text{A}$

**Note 1:** Boldface limits apply at temperature extremes.

**Note 2:** Unless otherwise specified all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V_{\text{IN}} = 6\text{V}$ ,  $I_L = 100\mu\text{A}$  and  $C_L = 1\mu\text{F}$ . Additional conditions for the 8-pin versions are Feedback tied to 5V Tap and Output tied to Output Sense ( $V_{\text{OUT}} = 5\text{V}$ ) and  $V_{\text{SHUTDOWN}} \leq 0.8\text{V}$ .

**Note 3:** Guaranteed and 100% production tested.

**Note 4:** Guaranteed but not 100% production tested. These limits are not used to calculate outgoing AQL levels.

**Note 5:** Dropout voltage is defined as the input to output differential at which the output voltage drops 100mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.

**Note 6:** Comparator thresholds are expressed in terms of a voltage differential at the Feedback terminal below the nominal reference voltage measured at 6V input. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain  $= V_{\text{OUT}}/V_{\text{REF}} = (R1 + R2)/R2$ . For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by  $95\text{mV} \times 5\text{V}/1.235\text{V} = 384\text{mV}$ . Thresholds remain constant as a percent of  $V_{\text{OUT}}$  as  $V_{\text{OUT}}$  is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.

**Note 7:**  $V_{\text{REF}} \leq V_{\text{OUT}} \leq (V_{\text{IN}} - 1\text{V})$ ,  $2.3\text{V} \leq V_{\text{IN}} \leq 30\text{V}$ ,  $100\mu\text{A} < I_L \leq 100\text{mA}$ ,  $T_J \leq T_{\text{JMAX}}$ .

**Note 8:** The junction-to-ambient thermal resistance of the TO-92 package is  $180^\circ\text{C/W}$  with 0.4" leads and  $160^\circ\text{C/W}$  with 0.25" leads to a PC board. The thermal resistance of the 8-pin DIP package is  $105^\circ\text{C/W}$  junction-to-ambient when soldered directly to a PC board. Junction-to-ambient thermal resistance for the SOIC (M) package is  $160^\circ\text{C/W}$ .

**Note 9:** May exceed input supply voltage.

**Note 10:** When used in dual-supply systems where the output terminal sees loads returned to a negative supply, the output voltage should be diode-clamped to ground.

**Note 11:**  $V_{\text{SHUTDOWN}} \geq 2\text{V}$ ,  $V_{\text{IN}} \leq 30\text{V}$ ,  $V_{\text{OUT}} = 0$ , with Feedback pin tied to 5V Tap.

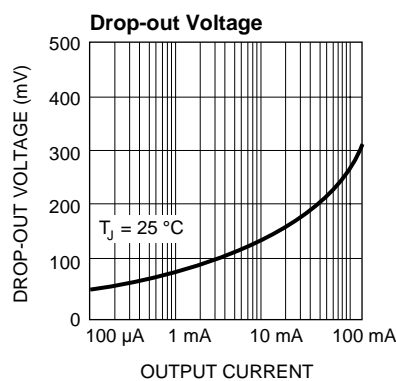
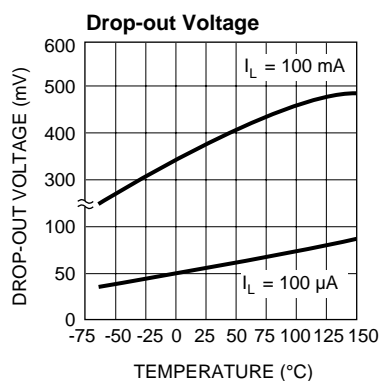
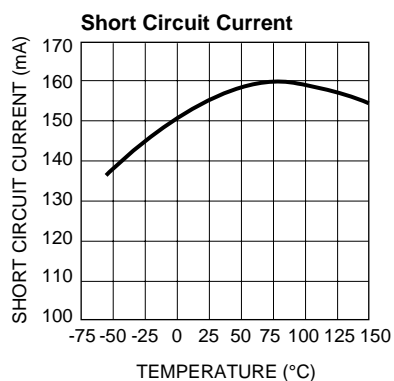
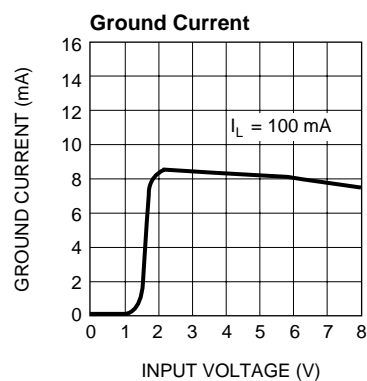
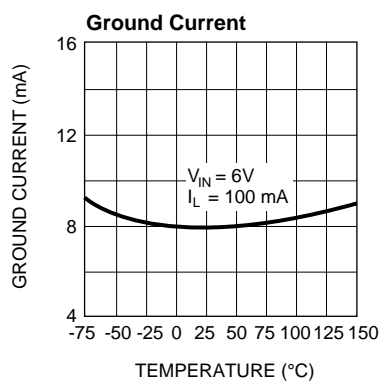
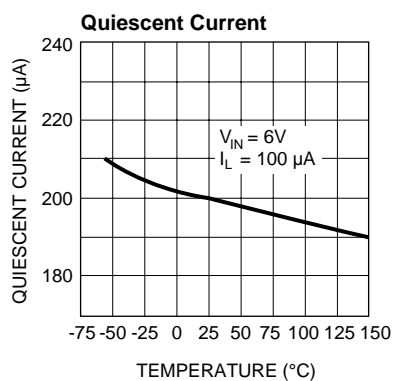
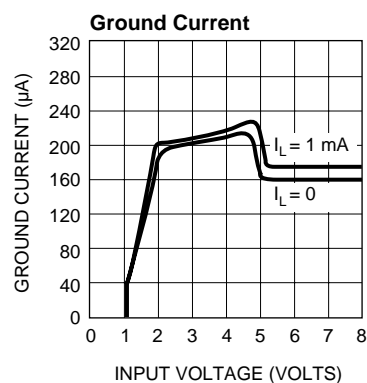
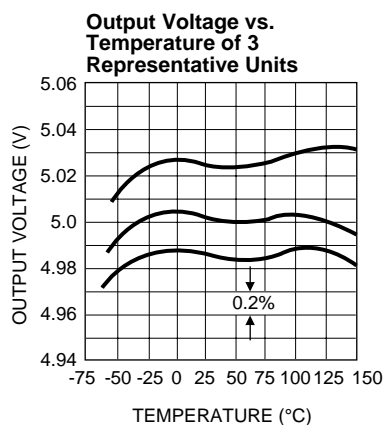
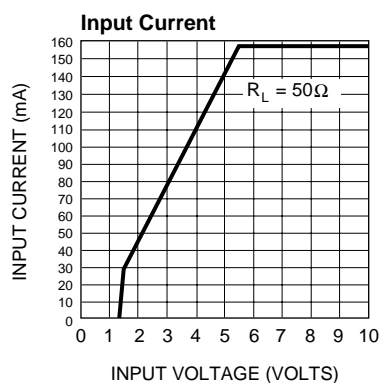
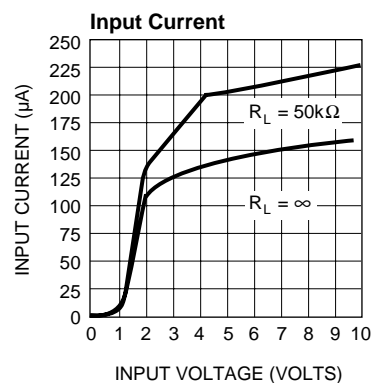
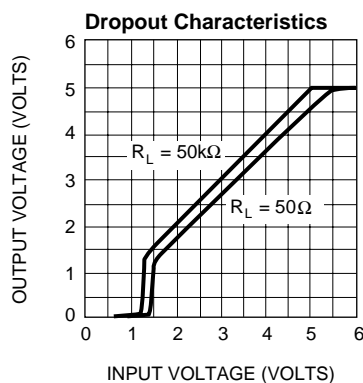
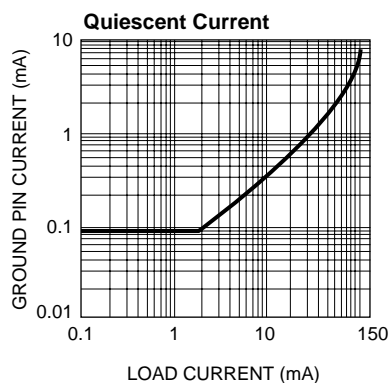
**Note 12:** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

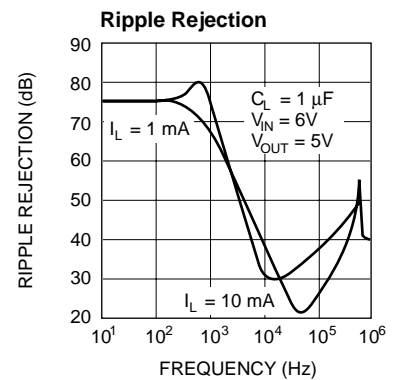
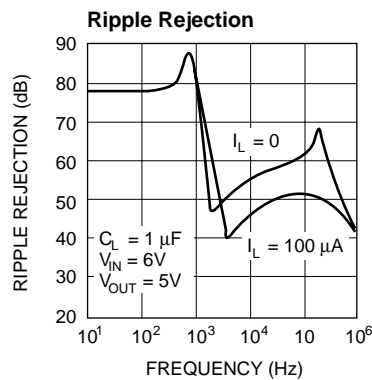
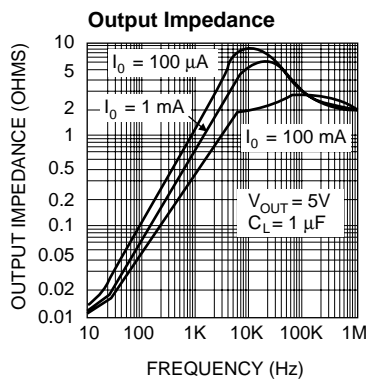
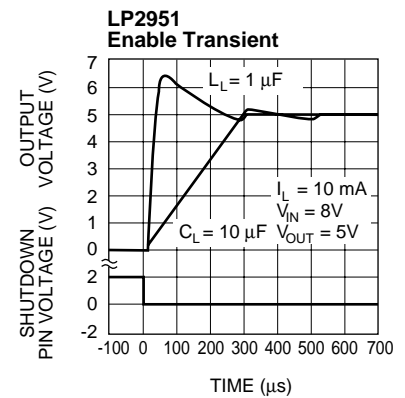
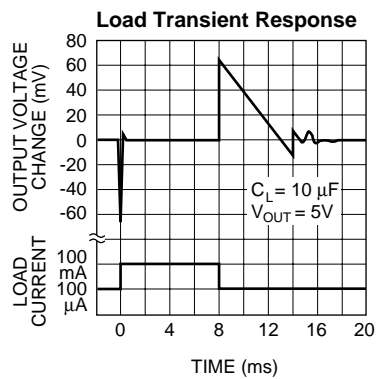
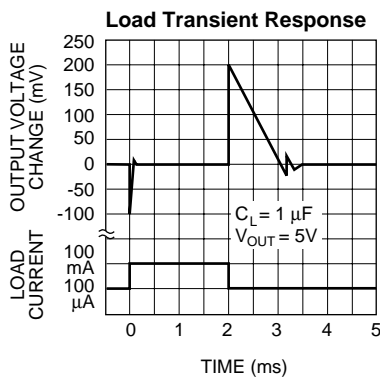
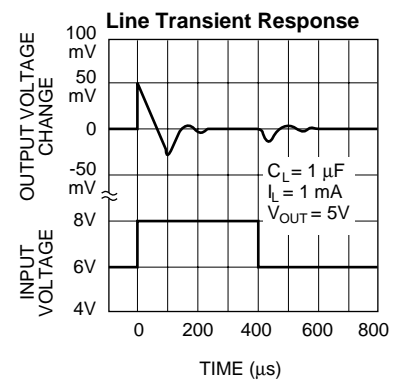
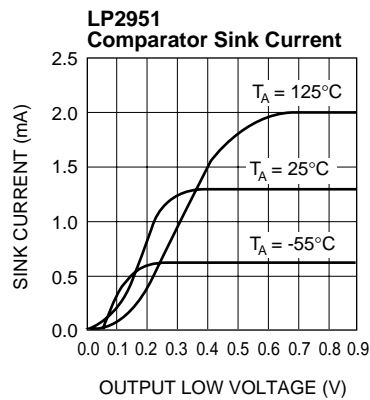
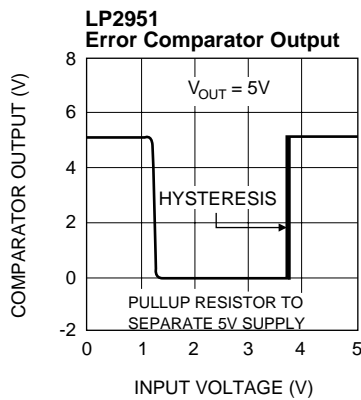
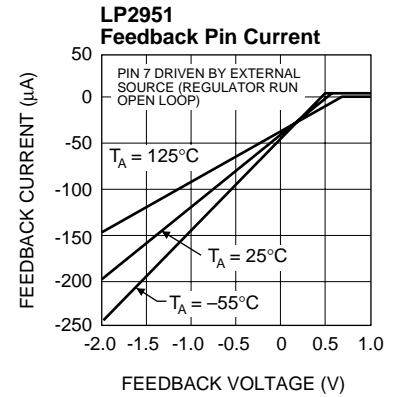
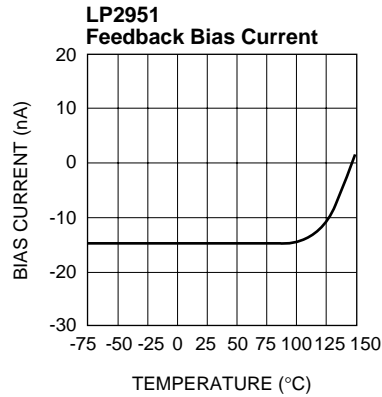
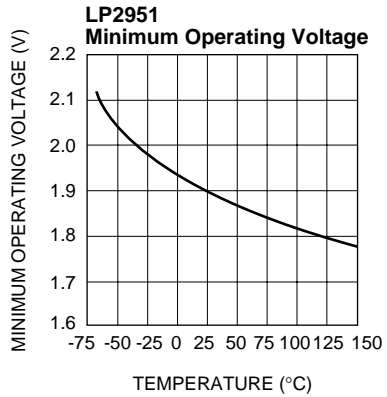
**Note 13:** Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50mA load pulse at  $V_{\text{IN}} = 30\text{V}$  (1.25W pulse) for  $t = 10\text{ms}$ .

**Note 14:** Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered in the specification for thermal regulation.

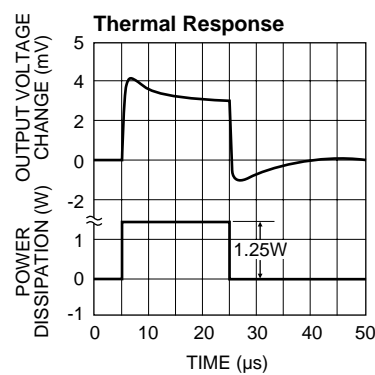
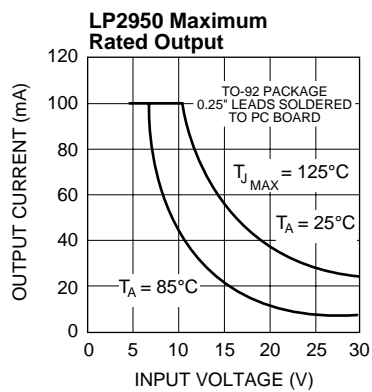
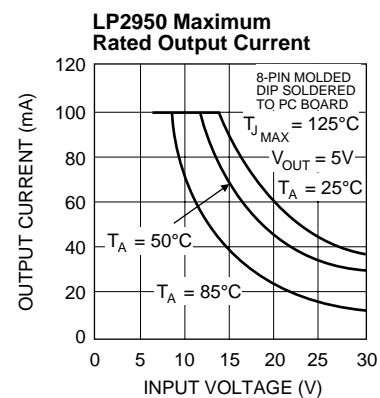
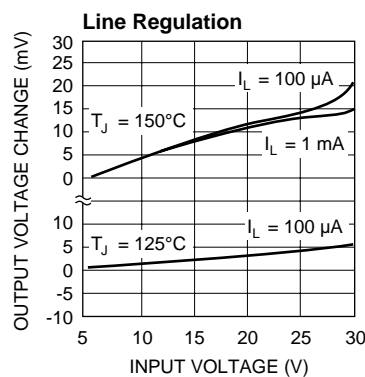
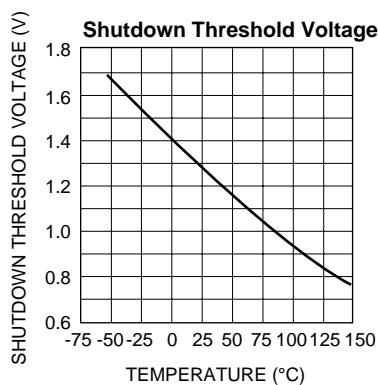
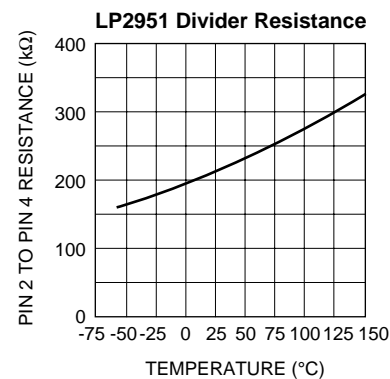
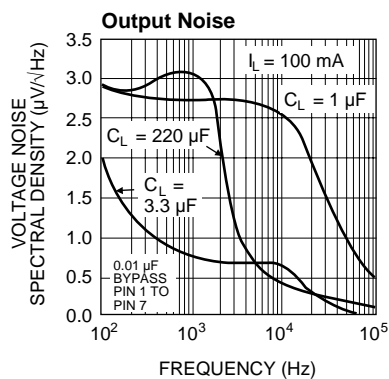
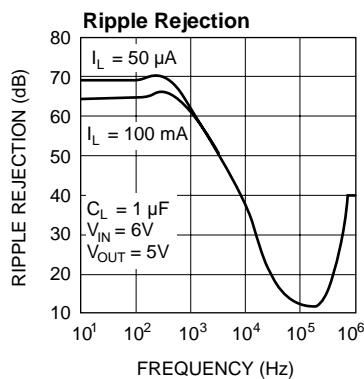
**Note 15:** Line regulation for the LP2951 is tested at  $150^\circ\text{C}$  for  $I_L = 1\text{mA}$ . For  $I_L = 100\mu\text{A}$  and  $T_J = 125^\circ\text{C}$ , line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.

# Typical Performance Characteristics





# Typical Performance Characteristics (Continued)



## Applications Information

### External Capacitors

A 1.0 $\mu$ F (or greater) capacitor is required between the LP2950/LP2951 output and ground to prevent oscillations due to instability. Most types of tantalum or aluminum electrolytics will be adequate; film types will work, but are costly and therefore not recommended. Many aluminum electrolytics have electrolytes that freeze at about  $-30^{\circ}\text{C}$ , so solid tantalum capacitors are recommended for operation below  $-25^{\circ}\text{C}$ . The important parameters of the capacitor are an effective series resistance of about  $5\Omega$  or less and a resonant frequency above 500kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.33 $\mu$ F for current below 10mA or 0.1 $\mu$ F for currents below 1mA. Using the 8-Pin versions at voltages below 5V runs the error amplifier at lower gains so that more output capacitance is needed. For the worst-case situation of a 100mA load at 1.23V output (Output shorted to Feedback) a 3.3 $\mu$ F (or greater) capacitor should be used.

The LP2950 will remain stable and in regulation with no load in addition to the internal voltage divider, unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the LP2951 version with external resistors, a minimum load of 1 $\mu$ A is recommended.

A 0.1 $\mu$ F capacitor should be placed from the LP2950/LP2951 input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

Stray capacitance to the LP2951 Feedback terminal (pin 7) can cause instability. This may especially be a problem when using high value external resistors to set the output voltage. Adding a 100pF capacitor between Output and Feedback and increasing the output capacitor to at least 3.3 $\mu$ F will remedy this.

### Error Detection Comparator Output

A logic low output will be produced by the comparator whenever the LP2951 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60mV divided by the 1.235V reference voltage. (Refer to the block diagram on Page 1). This trip level remains "5% below normal" regardless of the programmed output voltage of the LP2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting.

Figure 1 is a timing diagram depicting the  $\overline{\text{ERROR}}$  signal and the regulated output voltage as the LP2951 input is ramped up and down. The  $\overline{\text{ERROR}}$  signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at

which  $V_{\text{OUT}} = 4.75\text{V}$ ). Since the LP2951's dropout voltage is load-dependent (see curve in Typical Performance Characteristics), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point (approximately 4.75V) does not vary with load.

The error comparator has an open-collector output which requires an external pull-up resistor. Depending on system requirements, this resistor may be returned to the 5V output or some other supply voltage. In determining a value for this resistor, note that while the output is rated to sink 400 $\mu$ A, this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1M $\Omega$ . The resistor is not required if this output is unused.

### Programming the Output Voltage (LP2951)

The LP2951 may be pin-strapped for 5V using its internal voltage divider by tying Pin 1 (output) to Pin 2 (SENSE) and Pin 7 (FEEDBACK) to Pin 6 (5V TAP). Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. An external pair of resistors is required, as shown in Figure 2.

The complete equation for the output voltage is

$$V_{\text{OUT}} = V_{\text{REF}} \times \{ 1 + R_1/R_2 \} + I_{\text{FB}} R_2$$

where  $V_{\text{REF}}$  is the nominal 1.235 reference voltage and  $I_{\text{FB}}$  is the feedback pin bias current, nominally 20 nA. The minimum recommended load current of 1  $\mu$ A forces an upper limit of 1.2 M $\Omega$  on the value of  $R_2$ , if the regulator must work with no load (a condition often found in CMOS in standby),  $I_{\text{FB}}$  will produce a 2% typical error in  $V_{\text{OUT}}$  which may be eliminated at room temperature by trimming  $R_1$ . For better accuracy, choosing  $R_2 = 100\text{k}\Omega$  reduces this error to 0.17% while increasing the resistor program current to 12 $\mu$ A. Since the LP2951 typically draws 60  $\mu$ A at no load with Pin 2 open-circuited, this is a small price to pay.

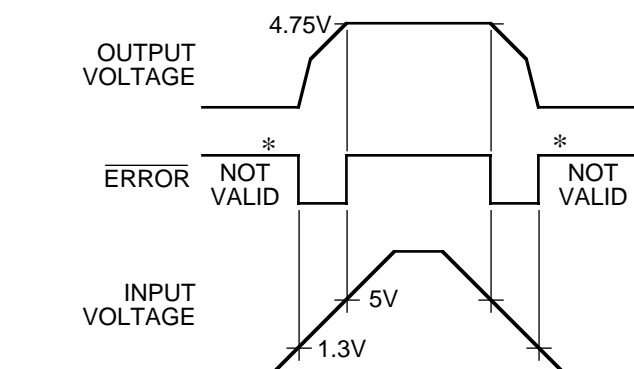
### Reducing Output Noise

In reference applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is the only method by which noise can be reduced on the 3 lead LP2950 and is relatively inefficient, as increasing the capacitor from 1 $\mu$ F to 220 $\mu$ F only decreases the noise from 430 $\mu$ V to 160 $\mu$ V rms for a 100kHz bandwidth at 5V output.

Noise can be reduced fourfold by a bypass capacitor across  $R_1$ , since it reduces the high frequency gain from 4 to unity. Pick

$$C_{\text{BYPASS}} \cong \frac{1}{2\pi R_1 \cdot 200 \text{ Hz}}$$

or about 0.01  $\mu$ F. When doing this, the output capacitor must be increased to 3.3  $\mu$ F to maintain stability. These changes reduce the output noise from 430 $\mu$ V to 100 $\mu$ V rms for a 100kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.



\* SEE APPLICATIONS INFORMATION

Figure 1. ERROR Output Timing

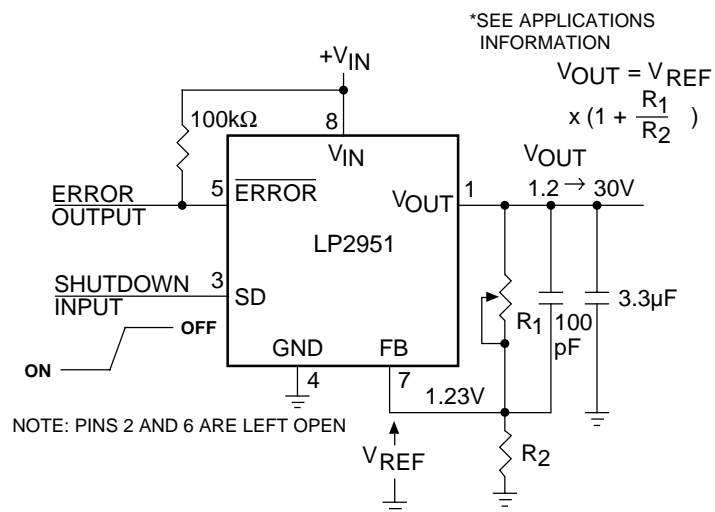
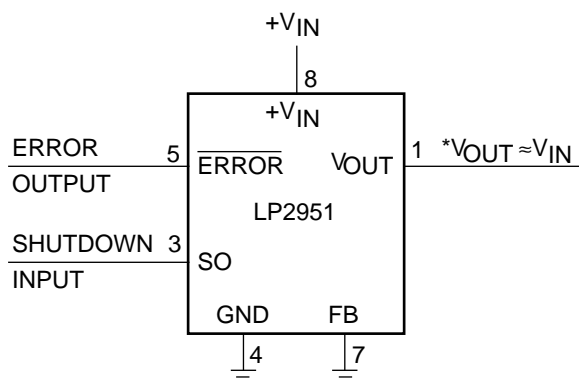


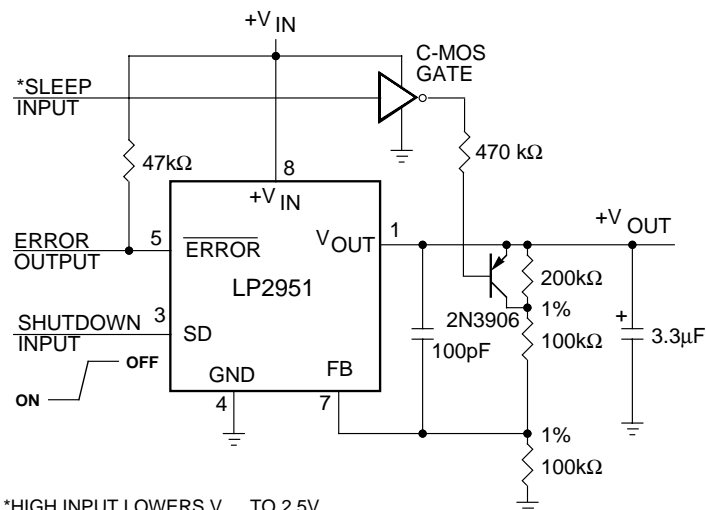
Figure 2. Adjustable Regulator

## Typical Applications



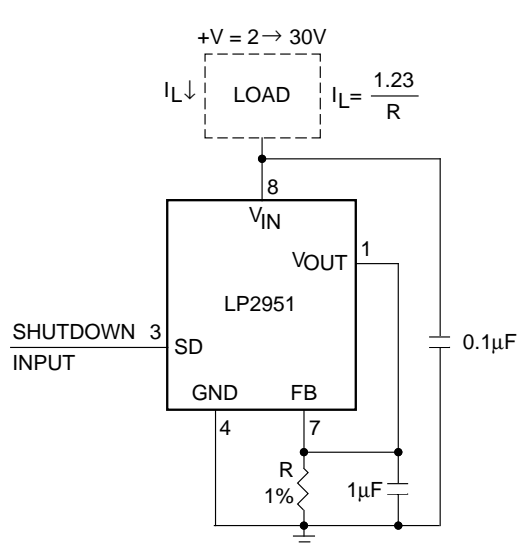
\*MINIMUM INPUT-OUTPUT VOLTAGE RANGES FROM 40mV TO 400mV, DEPENDING ON LOAD CURRENT. CURRENT LIMIT IS TYPICALLY 160mA.

Wide Input Voltage Range Current Limiter

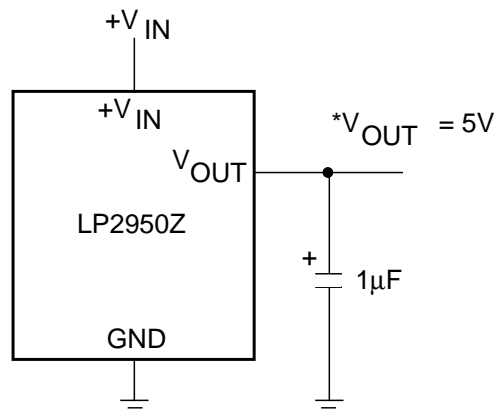


\*HIGH INPUT LOWERS  $V_{OUT}$  TO 2.5V

5 V Regulator with 2.5 V Sleep Function

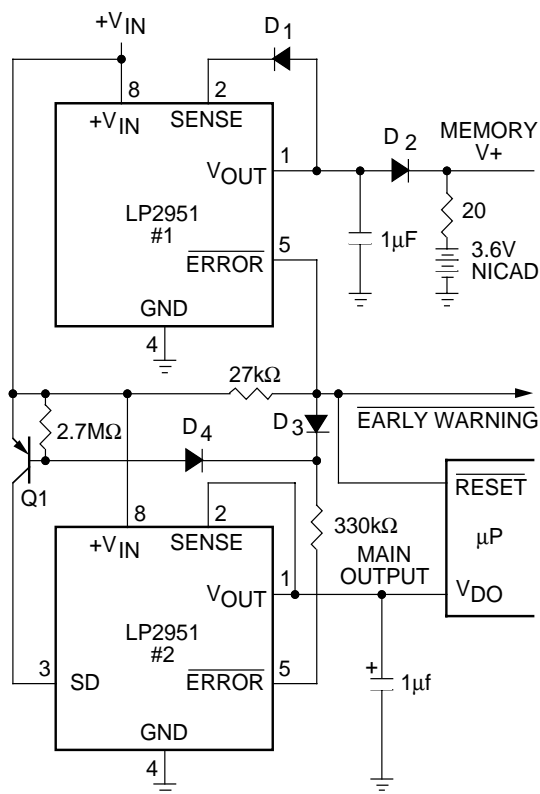


Low Drift Current Source



5 Volt Current Limiter

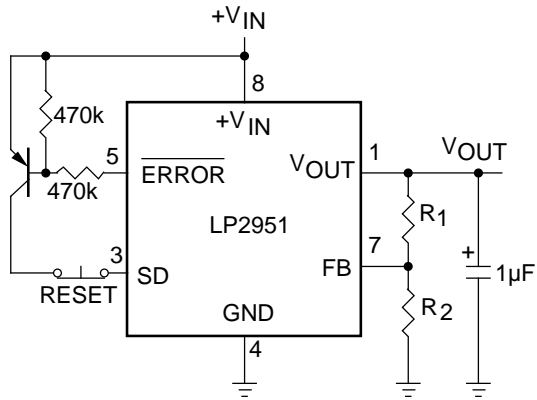
\* MINIMUM INPUT-OUTPUT VOLTAGE RANGES FROM 40mV TO 400mV, DEPENDING ON LOAD CURRENT.



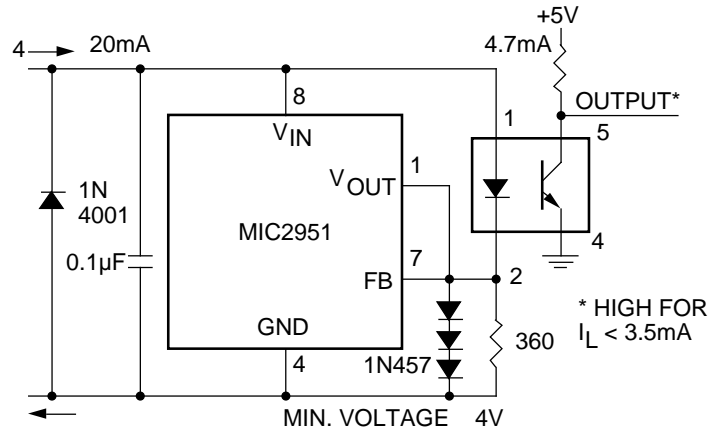
Regulator with Early Warning and Auxiliary Output

- EARLY WARNING FLAG ON LOW INPUT VOLTAGE
- MAIN OUTPUT LATCHES OFF AT LOWER INPUT VOLTAGES
- BATTERY BACKUP ON AUXILIARY OUTPUT

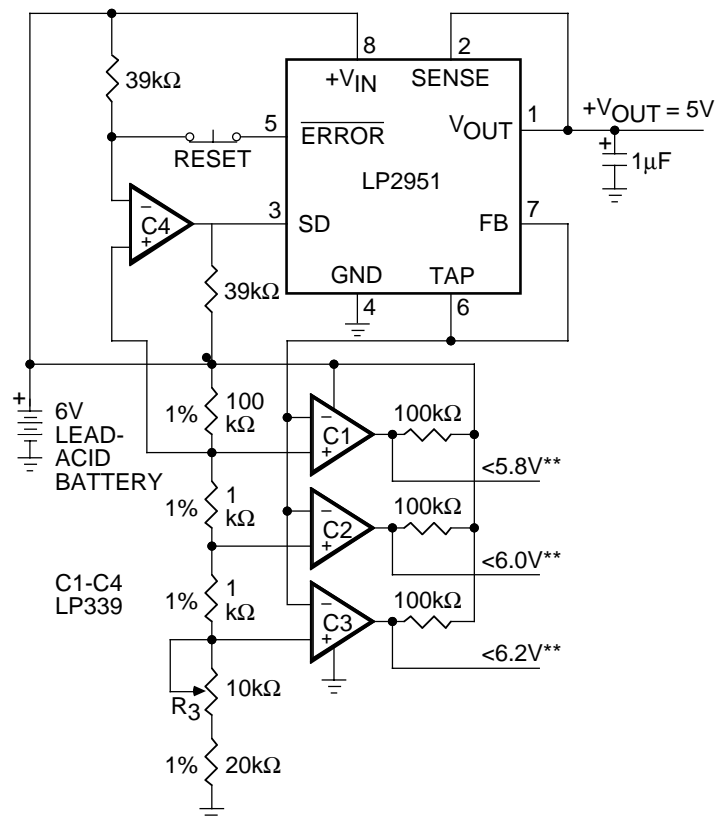
OPERATION: REG. #1'S  $V_{OUT}$  IS PROGRAMMED ONE DIODE DROP ABOVE 5 V. ITS ERROR FLAG BECOMES ACTIVE WHEN  $V_{IN} \leq 5.7$  V. WHEN  $V_{IN}$  DROPS BELOW 5.3 V, THE ERROR FLAG OF REG. #2 BECOMES ACTIVE AND VIA Q1 LATCHES THE MAIN OUTPUT OFF. WHEN  $V_{IN}$  AGAIN EXCEEDS 5.7 V REG. #1 IS BACK IN REGULATION AND THE EARLY WARNING SIGNAL RISES, UNLATCHING REG. #2 VIA D3.



Latch Off When Error Flag Occurs

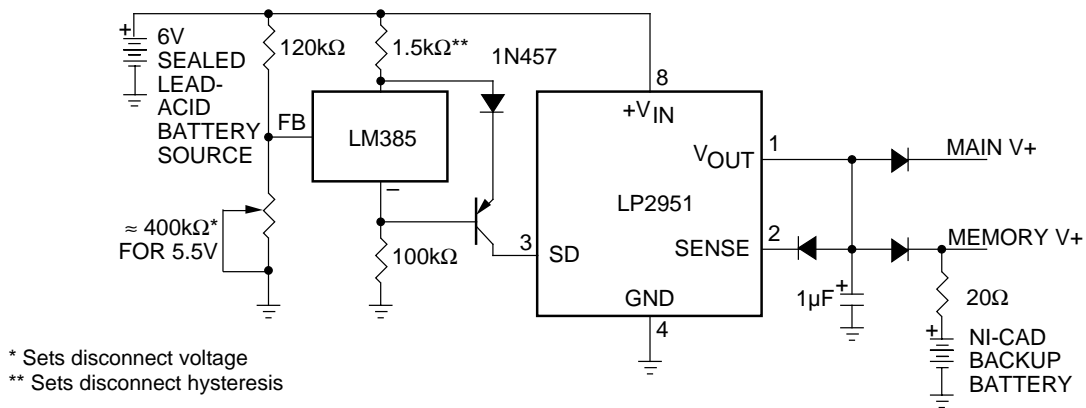


Open Circuit Detector for 4mA to 20mA Current Loop



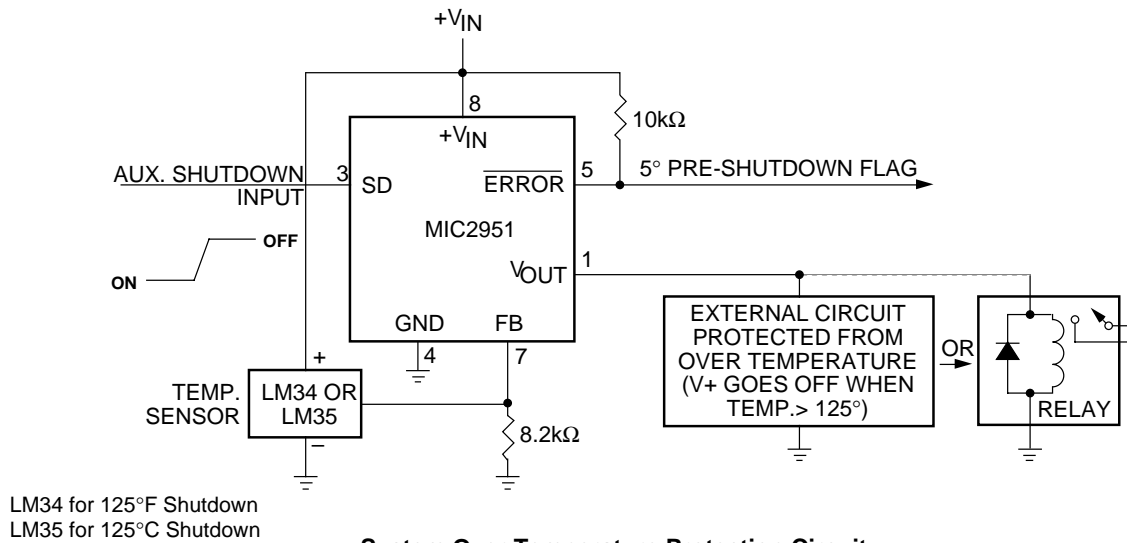
\*OPTIONAL LATCH OFF WHEN DROP OUT OCCURS. ADJUST R3 FOR C2 SWITCHING WHEN  $V_{IN}$  IS 6.0V  
 \*\*OUTPUTS GO LOW WHEN  $V_{IN}$  DROPS BELOW DESIGNATED THRESHOLDS.

Regulator with State-of-Charge Indicator



### Low Battery Disconnect

For values shown, Regulator shuts down when  $V_{IN} < 5.5\text{ V}$  and turns on again at  $6.0\text{ V}$ . Current drain in disconnected mode is  $150\mu\text{A}$ .



### System Over Temperature Protection Circuit

