



The Infinite Bandwidth Company™

MIC5207

180mA Low-Noise LDO Regulator

General Description

The MIC5207 is an efficient linear voltage regulator with ultra-low-noise output, very low dropout voltage (typically 17mV at light loads and 165mV at 150mA), and very low ground current (720μA at 100mA output). The MIC5207 offers better than 3% initial accuracy.

Designed especially for hand-held, battery-powered devices, the MIC5207 includes a CMOS or TTL compatible enable/shutdown control input. When shutdown, power consumption drops nearly to zero.

Key MIC5207 features include a reference bypass pin to improve its already low-noise performance, reversed-battery protection, current limiting, and overtemperature shutdown.

The MIC5207 is available in fixed and adjustable output voltage versions in a small SOT-23-5 package. Contact Micrel for details.

For low-dropout regulators that are stable with ceramic output capacitors, see the μCap MIC5245/6/7 family.

Features

- Ultra-low-noise output
- High output voltage accuracy
- Guaranteed 180mA output
- Low quiescent current
- Low dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Current and thermal limiting
- Reverse-battery protection
- "Zero" off-mode current
- Logic-controlled electronic enable

Applications

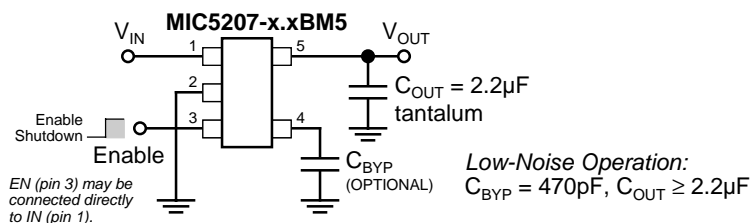
- Cellular telephones
- Laptop, notebook, and palmtop computers
- Battery-powered equipment
- PCMCIA V_{CC} and V_{PP} regulation/switching
- Consumer/personal electronics
- SMPS post-regulator/dc-to-dc modules
- High-efficiency linear power supplies

Ordering Information

Part Number*	Marking	Voltage	Junction Temp. Range	Package
MIC5207BM5	LEAA	Adj	-40°C to +125°C	SOT-23-5
MIC5207-1.8BM5	LE18	1.8	0°C to +125°C	SOT-23-5
MIC5207-2.5BM5	LE25	2.5	-40°C to +125°C	SOT-23-5
MIC5207-2.8BM5	LE28	2.8	-40°C to +125°C	SOT-23-5
MIC5207-3.0BM5	LE30	3.0	-40°C to +125°C	SOT-23-5
MIC5207-3.3BM5	LE33	3.3	-40°C to +125°C	SOT-23-5
MIC5207-3.6BM5	LE36	3.6	-40°C to +125°C	SOT-23-5
MIC5207-3.8BM5	LE38	3.8	-40°C to +125°C	SOT-23-5
MIC5207-4.0BM5	LE40	4.0	-40°C to +125°C	SOT-23-5
MIC5207-5.0BM5	LE50	5.0	-40°C to +125°C	SOT-23-5
MIC5207-3.3BZ	—	3.3	-40°C to +125°C	TO-92

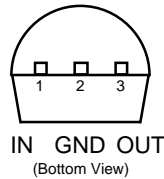
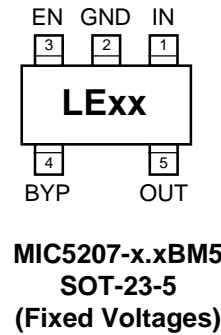
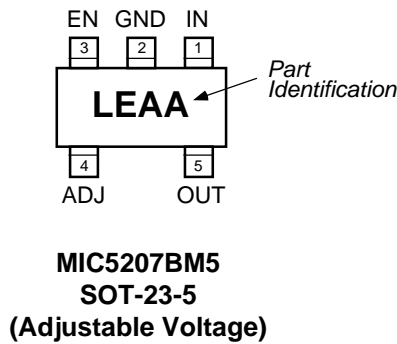
* Other voltages available. Contact Micrel Marketing for information.

Typical Application



Battery-Powered Regulator Application

Pin Configuration



Pin Description

Pin No. SOT-23-5	Pin No. TO-92	Pin Name	Pin Function
1	1	IN	Supply Input
2	2	GND	Ground
3		EN	Enable/Shutdown (Input): CMOS compatible input. Logic high = enable, logic low or open = shutdown.
4 (fix)		BYP	Reference Bypass: Connect external 470pF capacitor to GND to reduce output noise. May be left open. For 1.8V or 2.5V operation, see "Applications Information."
4 (adj)		ADJ	Adjust (Input): Adjustable regulator feedback input. Connect to resistor voltage divider.
5	3	OUT	Regulator Output

Absolute Maximum Ratings (Note 1)

Supply Input Voltage (V_{IN}) -20V to +20V
 Enable Input Voltage (V_{EN}) -20V to +20V
 Power Dissipation (P_D) Internally Limited, **Note 3**
 Lead Temperature (soldering, 5 sec.) 260°C
 Junction Temperature (T_J)
 all except 1.8V -40°C to +125°C
 1.8V only 0°C to +125°C
 Storage Temperature (T_S) -65°C to +150°C

Operating Ratings (Note 2)

Input Voltage (V_{IN}) +2.5V to +16V
 Enable Input Voltage (V_{EN}) 0V to V_{IN}
 Junction Temperature (T_J)
 all except 1.8V -40°C to +125°C
 1.8V only 0°C to +125°C
 Thermal Resistance (θ_{JA}) **Note 3**

Electrical Characteristics

$V_{IN} = V_{OUT} + 1V$; $I_L = 100\mu A$; $C_L = 1.0\mu F$; $V_{EN} \geq 2.0V$; $T_J = 25^\circ C$, **bold** values indicate $-40^\circ C \leq T_J \leq +125^\circ C$ except $0^\circ C \leq T_J \leq +125^\circ C$ for 1.8V version; unless noted.

Symbol	Parameter	Conditions	Min	Typical	Max	Units
V_O	Output Voltage Accuracy	variation from specified V_{OUT}	-3 -4		3 4	% %
$\Delta V_O / \Delta T$	Output Voltage Temperature Coefficient	Note 4		40		ppm/ $^\circ C$
$\Delta V_O / V_O$	Line Regulation	$V_{IN} = V_{OUT} + 1V$ to 16V		0.005	0.05 0.10	%/V %/V
$\Delta V_O / V_O$	Load Regulation	$I_L = 0.1mA$ to 150mA, Note 5		0.05	0.5 0.7	% %
$V_{IN} - V_O$	Dropout Voltage, Note 6	$I_L = 100\mu A$ $I_L = 50mA$ $I_L = 100mA$ $I_L = 150mA$		17 115 140 165	60 80 175 250 280 325 300 400	mV mV mV mV mV mV mV mV
I_{GND}	Quiescent Current	$V_{EN} \leq 0.4V$ (shutdown) $V_{EN} \leq 0.18V$ (shutdown)		0.01	1 5	μA μA
I_{GND}	Ground Pin Current, Note 7	$V_{EN} \geq 2.0V$, $I_L = 100\mu A$ $I_L = 50mA$ $I_L = 100mA$ $I_L = 150mA$		80 350 720 1800	130 170 650 900 1100 2000 2500 3000	μA μA μA μA μA μA μA μA
PSRR	Ripple Rejection			75		dB
I_{LIMIT}	Current Limit	$V_{OUT} = 0V$		320	500	mA
$\Delta V_O / \Delta P_D$	Thermal Regulation	Note 8		0.05		%/W
e_{no}	Output Noise	$I_L = 50mA$, $C_L = 2.2\mu F$, 470pF from BYP to GND		260		nV \sqrt{Hz}

ENABLE Input

V_{IL}	Enable Input Logic-Low Voltage	regulator shutdown			0.4 0.18	V V
V_{IH}	Enable Input Logic-High Voltage	regulator enabled	2.0			V
I_{IL}	Enable Input Current	$V_{IL} \leq 0.4V$ $V_{IL} \leq 0.18V$		0.01	-1 -2	μA μA
I_{IH}		$V_{IH} \geq 2.0V$ $V_{IH} \geq 2.0V$		5	20 25	μA μA

Note 1. Exceeding the absolute maximum rating may damage the device.

Note 2. The device is not guaranteed to function outside its operating rating.

Note 3: The maximum allowable power dissipation at any T_A (ambient temperature) is $P_{D(max)} = (T_{J(max)} - T_A) \div \theta_{JA}$. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. The θ_{JA} of the SOT-23-5 (M5) is $235^\circ C/W$ and the TO-92 (Z) is $180^\circ C/W$ (0.4" leads) or $160^\circ C/W$ (0.25" leads) soldered to a PC board. See "Thermal Considerations."

Note 4: Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

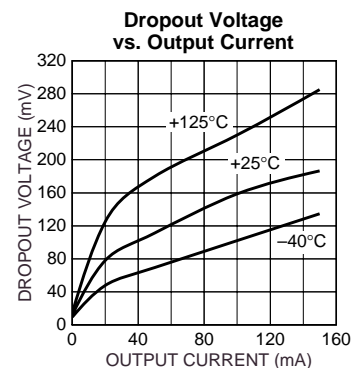
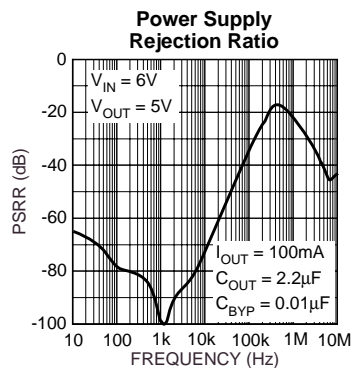
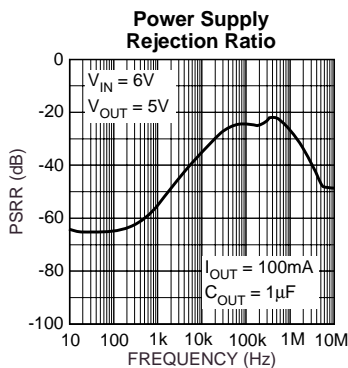
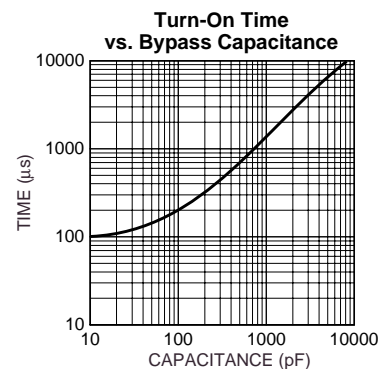
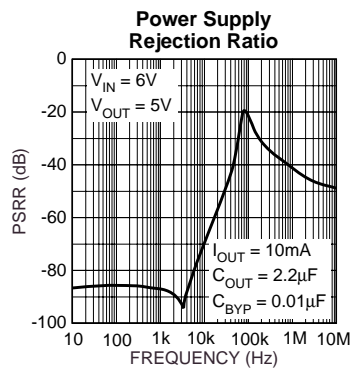
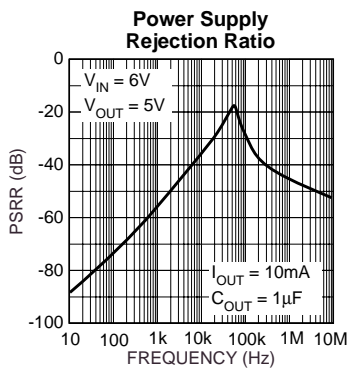
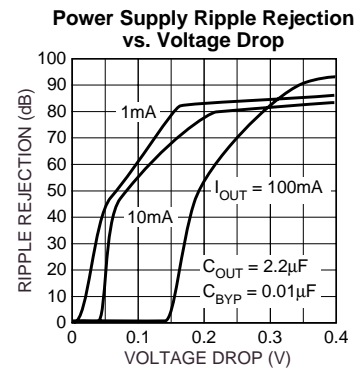
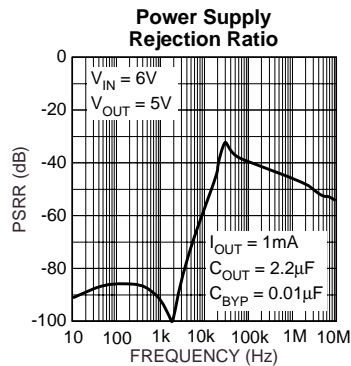
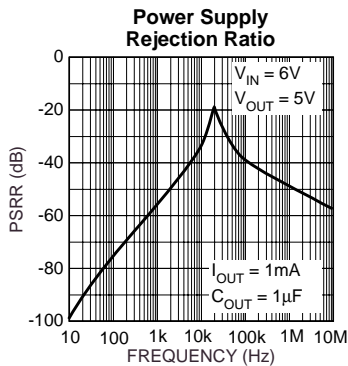
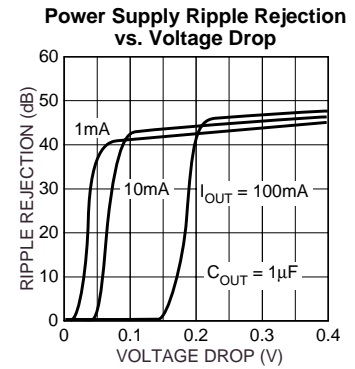
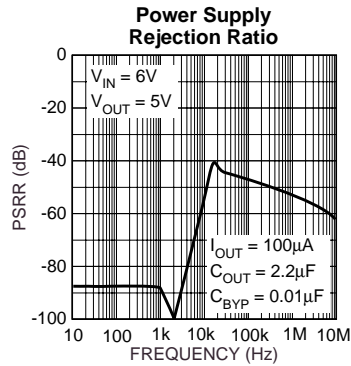
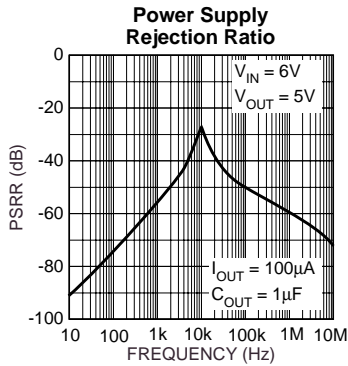
Note 5: Regulation is measured at constant junction temperature using low duty cycle pulse testing. Parts are tested for load regulation in the load range from 0.1mA to 180mA. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

Note 6: Dropout voltage is the input to output differential at which the output voltage drops 2% below its nominal value measured at 1V differential.

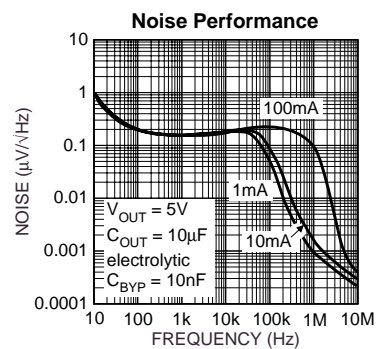
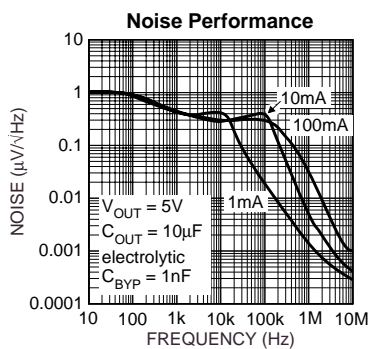
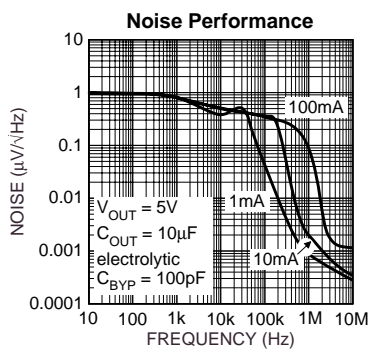
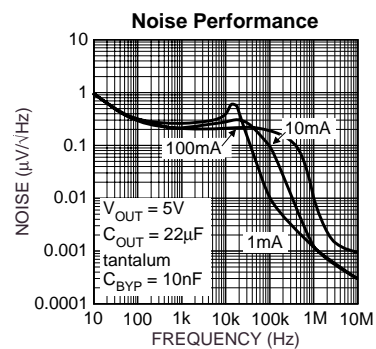
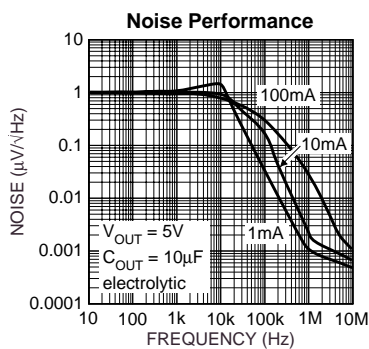
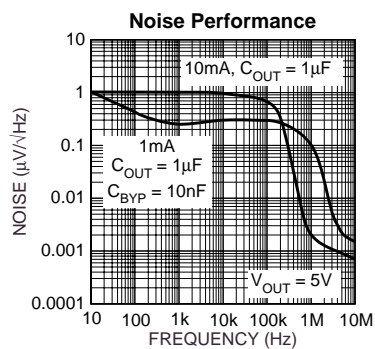
Note 7: Ground pin current is the regulator quiescent current plus pass transistor base current. The total current drawn from the supply is the sum of the load current plus the ground pin current.

Note 8: Thermal regulation is defined as the change in output voltage at a time "t" after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 180mA load pulse at $V_{IN} = 16V$ for $t = 10ms$.

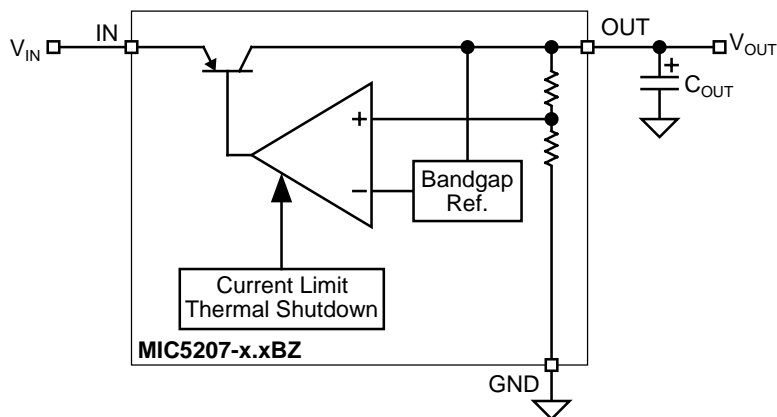
Typical Characteristics



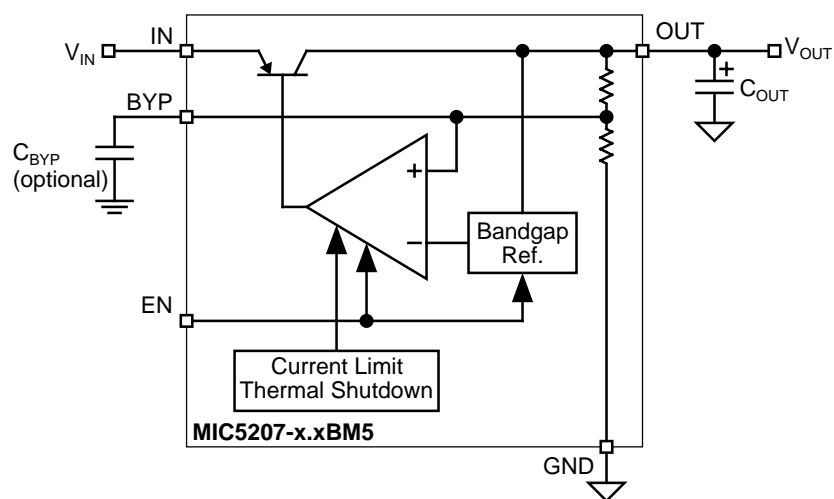
Typical Characteristics



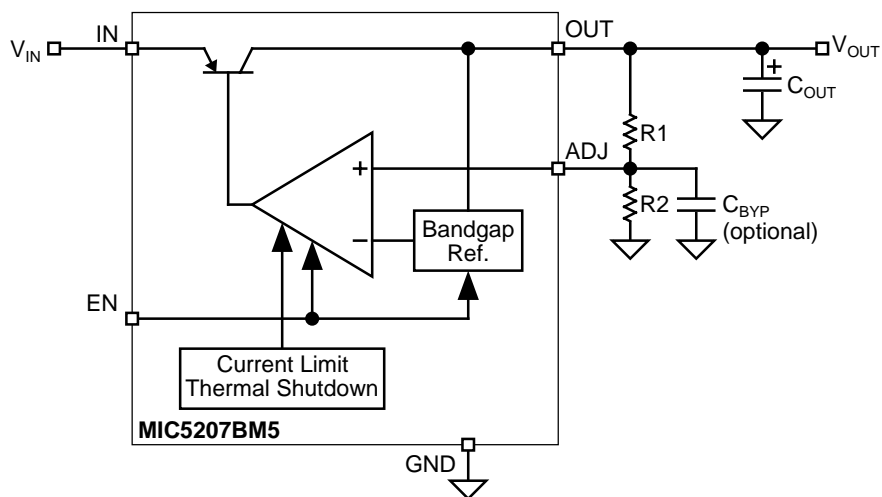
Block Diagrams



Low-Noise Fixed Regulator (TO-92 version only)



Ultra-Low-Noise Fixed Regulator



Ultra-Low-Noise Adjustable Regulator

Applications Information

Enable/Shutdown

Forcing EN (enable/shutdown) high (> 2V) enables the regulator. EN is compatible with CMOS logic gates.

If the enable/shutdown feature is not required, connect EN (pin 3) to IN (supply input, pin 1). See Figure 1.

Input Capacitor

A 1μF capacitor should be placed from IN to GND if there is more than 10 inches of wire between the input and the ac filter capacitor or if a battery is used as the input.

Reference Bypass Capacitor

BYP (reference bypass) is connected to the internal voltage reference. A 470pF capacitor (C_{BYP}) connected from BYP to GND quiets this reference, providing a significant reduction in output noise. C_{BYP} reduces the regulator phase margin; when using C_{BYP} , output capacitors of 2.2μF or greater are generally required to maintain stability.

The start-up speed of the MIC5207 is inversely proportional to the size of the reference bypass capacitor. Applications requiring a slow ramp-up of output voltage should consider larger values of C_{BYP} . Likewise, if rapid turn-on is necessary, consider omitting C_{BYP} .

If output noise is not a major concern, omit C_{BYP} and leave BYP open.

Output Capacitor

An output capacitor is required between OUT and GND to prevent oscillation. The minimum size of the output capacitor is dependent upon whether a reference bypass capacitor is used. 1.0μF minimum is recommended when C_{BYP} is not used (see Figure 2). 2.2μF minimum is recommended when C_{BYP} is 470pF (see Figure 1). Larger values improve the regulator's transient response. The output capacitor value may be increased without limit.

The output capacitor should have an ESR (effective series resistance) of about 5Ω or less and a resonant frequency above 1MHz. Ultra-low-ESR capacitors can cause a low amplitude oscillation on the output and/or underdamped transient response. Most tantalum or aluminum electrolytic capacitors are adequate; film types will work, but are more expensive. Since many aluminum electrolytics have electrolytes that freeze at about -30°C, solid tantalums are recommended for operation below -25°C.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.47μF for current below 10mA or 0.33μF for currents below 1mA.

No-Load Stability

The MIC5207 will remain stable and in regulation with no load (other than the internal voltage divider) unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications.

Thermal Considerations

The MIC5207 is designed to provide 180mA of continuous current in a very small package. Maximum power dissipation can be calculated based on the output current and the voltage

drop across the part. To determine the maximum power dissipation of the package, use the junction-to-ambient thermal resistance of the device and the following basic equation:

$$P_{D(max)} = \frac{(T_{J(max)} - T_A)}{\theta_{JA}}$$

$T_{J(max)}$ is the maximum junction temperature of the die, 125°C, and T_A is the ambient operating temperature. θ_{JA} is layout dependent; Table 1 shows examples of junction-to-ambient thermal resistance for the MIC5207.

Package	θ_{JA} Recommended Minimum Footprint	θ_{JA} 1" Square Copper Clad	θ_{JC}
SOT-23-5 (M5)	235°C/W	170°C/W	130°C/W

Table 1. SOT-23-5 Thermal Resistance

The actual power dissipation of the regulator circuit can be determined using the equation:

$$P_D = (V_{IN} - V_{OUT}) I_{OUT} + V_{IN} I_{GND}$$

Substituting $P_{D(max)}$ for P_D and solving for the operating conditions that are critical to the application will give the maximum operating conditions for the regulator circuit. For example, when operating the MIC5207-3.3BM5 at room temperature with a minimum footprint layout, the maximum input voltage for a set output current can be determined as follows:

$$P_{D(max)} = \frac{125^\circ\text{C} - 25^\circ\text{C}}{235}$$

$$P_{D(max)} = 425\text{mW}$$

The junction-to-ambient thermal resistance for the minimum footprint is 220°C/W, from Table 1. The maximum power dissipation must not be exceeded for proper operation. Using the output voltage of 3.3V and an output current of 150mA, the maximum input voltage can be determined. From the Electrical Characteristics table, the maximum ground current for 150mA output current is 3000μA or 3mA.

$$455\text{mW} = (V_{IN} - 3.3\text{V}) 150\text{mA} + V_{IN} \cdot 3\text{mA}$$

$$455\text{mW} = V_{IN} \cdot 150\text{mA} - 495\text{mW} + V_{IN} \cdot 3\text{mA}$$

$$920\text{mW} = V_{IN} \cdot 153\text{mA}$$

$$V_{IN(max)} = 6.01\text{V}$$

Therefore, a 3.3V application at 150mA of output current can accept a maximum input voltage of 6V in a SOT-23-5 package. For a full discussion of heat sinking and thermal effects on voltage regulators, refer to the Regulator Thermals section of Micrel's *Designing with Low-Dropout Voltage Regulators* handbook.

Low-Voltage Operation

The MIC5207-1.8 and MIC5207-2.5 require special consideration when used in voltage-sensitive systems. They may momentarily overshoot their nominal output voltages unless appropriate output and bypass capacitor values are chosen.

During regulator power up, the pass transistor is fully saturated for a short time, while the error amplifier and voltage reference are being powered up more slowly from the output

(see "Block Diagram"). Selecting larger output and bypass capacitors allows additional time for the error amplifier and reference to turn on and prevent overshoot.

To ensure that no overshoot is present when starting up into a light load (100µA), use a 4.7µF output capacitance and 470pF bypass capacitance. This slows the turn-on enough to allow the regulator to react and keep the output voltage from exceeding its nominal value. At heavier loads, use a 10µF output capacitance and 470pF bypass capacitance. Lower values of output and bypass capacitance can be used, depending on the sensitivity of the system.

Applications that can withstand some overshoot on the output of the regulator can reduce the output capacitor and/or reduce or eliminate the bypass capacitor. Applications that are not sensitive to overshoot due to power-on reset delays can use normal output and bypass capacitor configurations.

Please note the junction temperature range of the regulator at 1.8V output (fixed and adjustable) is 0°C to +125°C.

Fixed Regulator Applications

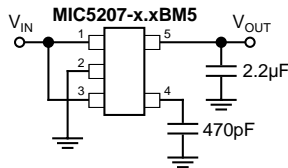


Figure 1. Ultra-Low-Noise Fixed Voltage Regulator

Figure 1 includes a 470pF capacitor for ultra-low-noise operation and shows EN (pin 3) connected to IN (pin 1) for an application where enable/shutdown is not required. $C_{OUT} = 2.2\mu\text{F}$ minimum.

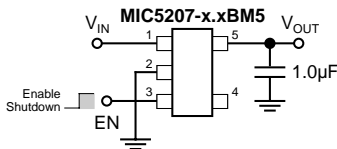


Figure 2. Low-Noise Fixed Voltage Regulator

Figure 2 is an example of a basic low-noise configuration. $C_{OUT} = 1\mu\text{F}$ minimum.

Adjustable Regulator Applications

The MIC5207BM5 can be adjusted to a specific output voltage by using two external resistors (figure 3). The resistors set the output voltage based on the following equation:

$$V_{OUT} = V_{REF} \left(1 + \frac{R_2}{R_1} \right), \quad V_{REF} = 1.242\text{V}$$

This equation is correct due to the configuration of the bandgap reference. The bandgap voltage is relative to the output, as seen in the block diagram. Traditional regulators normally have the reference voltage relative to ground; therefore, their equations are different from the equation for the MIC5207BM5.

Resistor values are not critical because ADJ (adjust) has a high input impedance, but for best results use resistors of 470kΩ or less. A capacitor from ADJ to ground provides greatly improved noise performance.

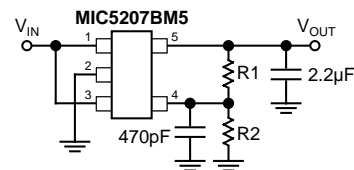


Figure 3. Ultra-Low-Noise Adjustable Voltage Regulator

Figure 3 includes the optional 470pF noise bypass capacitor from ADJ to GND to reduce output noise.

Dual-Supply Operation

When used in dual-supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.

USB Application

Figure 4 shows the MIC5207-3.3BZ (3-terminal, TO-92) in a USB application. Since the V_{BUS} supply may be greater than 10 inches from the regulator, a 1µF input capacitor is included.

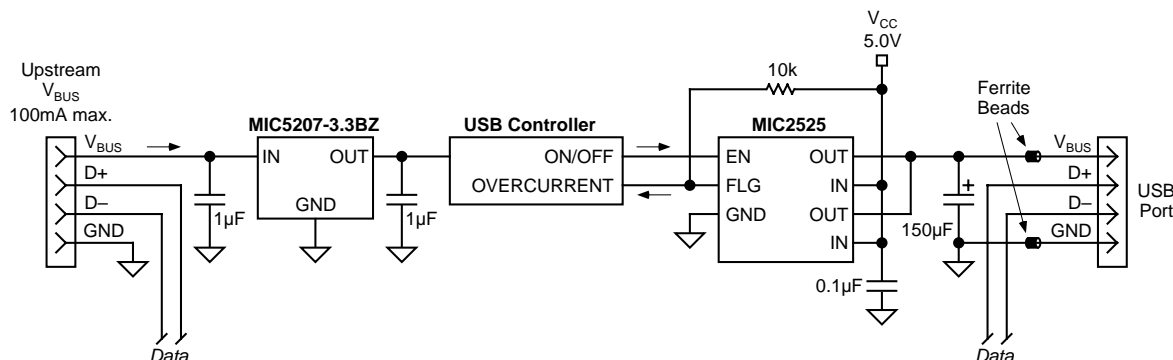
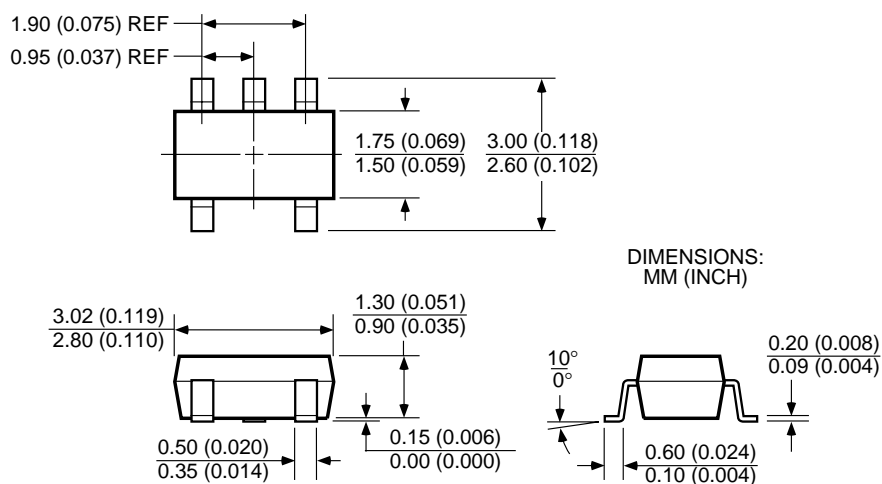
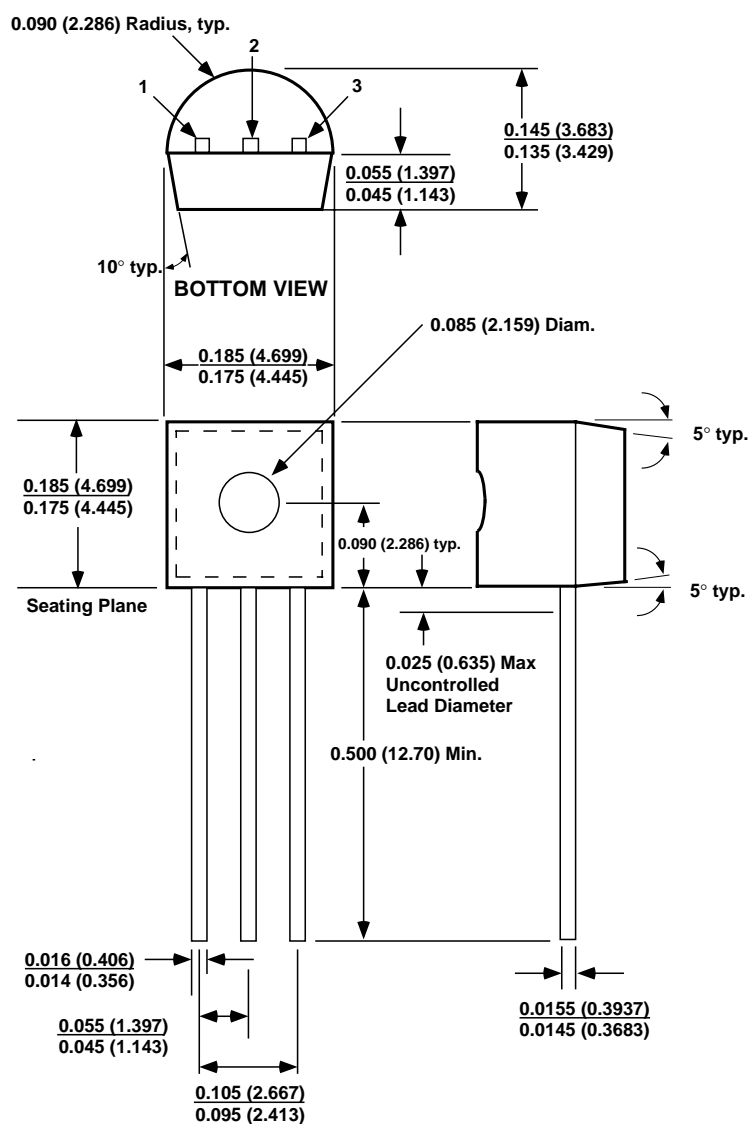


Figure 4. Single-Port Self-Powered Hub

Package Information



SOT-23-5 (M5)



TO-92 (Z)

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