



The Infinite Bandwidth Company™

MIC5236

Low Quiescent Current μ Cap LDO Regulator

Final Information

General Description

The MIC5236 is a low quiescent current, μ Cap low-dropout regulator. With a maximum operating input voltage of 30V and a quiescent current of 20 μ A, it is ideal for supplying keep-alive power in systems with high-voltage batteries.

Capable of 150mA output, the MIC5236 has a dropout voltage of only 300mV. It can also survive an input transient of -20V to +60V.

As a μ Cap LDO, the MIC5236 is stable with either a ceramic or a tantalum output capacitor. It only requires a 1.0 μ F output capacitor for stability.

The MIC5236 includes a logic compatible enable input and an undervoltage error flag indicator. Other features of the MIC5236 include thermal shutdown, current-limit, overvoltage shutdown, load-dump protection, reverse leakage protections, and reverse battery protection.

Available in the thermally enhanced SOP-8 and MSOP-8, the MIC5236 comes in fixed 2.5V, 3.0V, 3.3V, 5.0V, and adjustable voltages. For other output voltages, contact Micrel.

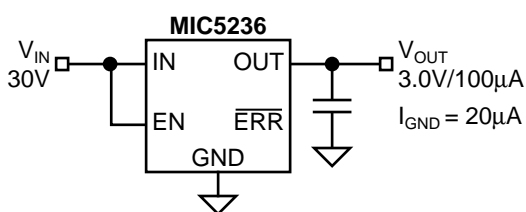
Features

- Ultra-low quiescent current ($I_Q = 20\mu\text{A}$ @ $I_O = 100\mu\text{A}$)
- Wide input range: 2.3V to 30V
- Low dropout:
 - 230mV @ 50mA;
 - 300mV @ 150mA
- Fixed 2.5V, 3.0V, 3.3V, 5.0V, and Adjustable outputs
- $\pm 1.0\%$ initial output accuracy
- Stable with ceramic or tantalum output capacitor
- Load dump protection: -20V to +60V input transient survivability
- Logic compatible enable input
- Low output flag indicator
- Overcurrent protection
- Thermal shutdown
- Reverse-leakage protection
- Reverse-battery protection
- High-power SOP-8 and MSOP-8

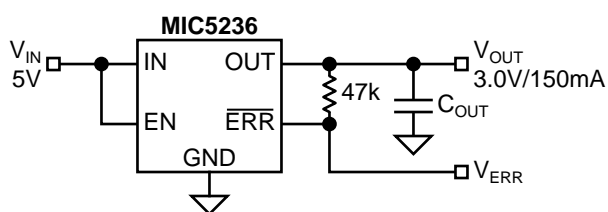
Applications

- Keep-alive supply in notebook and portable personal computers
- Logic supply from high-voltage batteries
- Automotive electronics
- Battery-powered systems

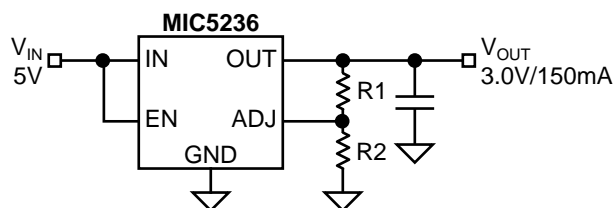
Typical Application



Regulator with Low I_O and Low I_Q



Regulator with Error Output



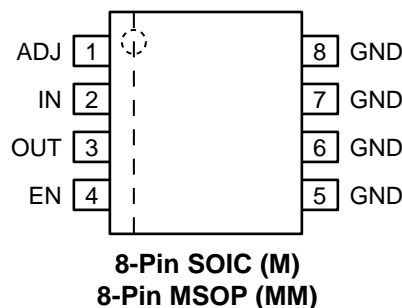
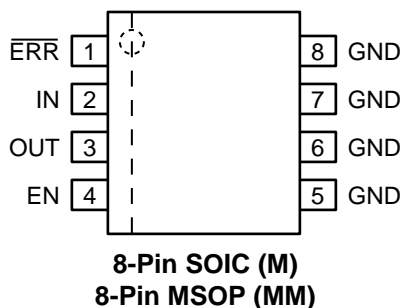
Regulator with Adjustable Output

Ordering Information

Part Number *	Voltage	Junction Temp. Range	Package
MIC5236-5.0BM	5.0V	−40°C to +125°C	8-lead SOIC
MIC5236-5.0BMM	5.0V	−40°C to +125°C	8-lead MSOP
MIC5236-3.3BM	3.3V	−40°C to +125°C	8-lead SOIC
MIC5236-3.3BMM	3.3V	−40°C to +125°C	8-lead MSOP
MIC5236-3.0BM	3.0V	−40°C to +125°C	8-lead SOIC
MIC5236-3.0BMM	3.0V	−40°C to +125°C	8-lead MSOP
MIC5236-2.5BM	2.5V	−40°C to +125°C	8-lead SOIC
MIC5236-2.5BMM	2.5V	−40°C to +125°C	8-lead MSOP
MIC5236BM	ADJ	−40°C to +125°C	8-lead SOIC
MIC5236BMM	ADJ	−40°C to +125°C	8-lead MSOP

*Contact factory regarding availability for voltages not listed

Pin Configuration



Pin Description

Pin Number	Pin Number	Pin Name	Pin Function
	1	/ERR	Error (Output): Open-collector output is active low when the output is out of regulation due to insufficient input voltage or excessive load. An external pull-up resistor is required.
1		ADJ	Adjustable Feedback Input. Connect to voltage divider network.
2	2	IN	Power supply input.
3	3	OUT	Regulated Output
4	4	EN	Enable (Input): Logic low = shutdown; logic high = enabled.
5–8	5–8	GND	Ground: Pins 5, 6, 7, and 8 are internally connected in common via the leadframe.

Absolute Maximum Ratings (Note 1)

Supply Voltage (V_{IN}), **Note 3** -20V to +60V
 Power Dissipation (P_D), **Note 4** Internally Limited
 Junction Temperature (T_J) +150°C
 Storage Temperature (T_S) -65°C to +150°C
 Lead Temperature (soldering, 5 sec.) 260°C
 ESD Rating, **Note 5**

Operating Ratings (Note 2)

Supply Voltage (V_{IN}) + 2.3V to +30V
 Junction Temperature (T_J) -40°C to +125°C
 Package Thermal Resistance
 MSOP (θ_{JA}) 80°C/W
 SOIC (θ_{JA}) 63°C/W

Electrical Characteristics

$V_{IN} = 6.0V$; $V_{EN} = 2.0V$; $C_{OUT} = 4.7\mu F$; $I_{OUT} = 100\mu A$; $T_J = 25^\circ C$, **bold** values indicate $-40^\circ C \leq T_J \leq +125^\circ C$; unless noted.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
V_{OUT}	Output Voltage Accuracy	variation from nominal V_{OUT}	-1 -2		1 +2	% %
$\Delta V_{OUT}/\Delta T$	Output Voltage Temperature Coefficient	Note 6		50		ppm/°C
$\Delta V_{OUT}/V_{OUT}$	Line Regulation	$V_{IN} = V_{OUT} + 1V$ to 30V		0.2	0.5 1.0	% %
$\Delta V_{OUT}/V_{OUT}$	Load Regulation	$I_{OUT} = 100\mu A$ to 50mA, Note 7		0.15	0.3 0.5	% %
		$I_{OUT} = 100\mu A$ to 150mA, Note 7		0.3	0.6 1.0	% %
ΔV	Dropout Voltage, Note 8	$I_{OUT} = 100\mu A$		50	100	mV
		$I_{OUT} = 50mA$		230	400	mV
		$I_{OUT} = 100mA$		270		mV
		$I_{OUT} = 150mA$		300	500	mV
I_{GND}	Ground Pin Current	$V_{EN} \geq 2.0V$, $I_{OUT} = 100\mu A$		20	30	μA
		$V_{EN} \geq 2.0V$, $I_{OUT} = 50mA$		0.5	0.8	mA
		$V_{EN} \geq 2.0V$, $I_{OUT} = 100mA$		1.5		mA
		$V_{EN} \geq 2.0V$, $I_{OUT} = 150mA$		2.8	4.0 5.0	mA mA
$I_{GND(SHDN)}$	Ground Pin in Shutdown	$V_{EN} \leq 0.6V$, $V_{IN} = 30V$		0.1	1	μA
I_{SC}	Short Circuit Current	$V_{OUT} = 0V$		260	350	mA
e_n	Output Noise	10Hz to 100kHz, $V_{OUT} = 3.0V$, $C_L = 1.0\mu F$		160		μV_{rms}

/ERR Output

V_{ERR}	Low Threshold	% of V_{OUT}	90	94		%
	High Threshold	% of V_{OUT}		95	98	%
V_{OL}	/ERR Output Low Voltage	$V_{IN} = V_{OUT(nom)} - 0.12V_{OUT}$, $I_{OL} = 200\mu A$		150	250 400	mV mV
I_{LEAK}	/ERR Output Leakage	$V_{OH} = 30V$		0.1	1 2	μA μA

Enable Input

V_{IL}	Input Low Voltage	regulator off			0.6	V
V_{IH}	Input High Voltage	regulator on	2.0			V

Symbol	Parameter	Conditions	Min	Typ	Max	Units
I_{IN}	Enable Input Current	$V_{EN} = 0.6V$, regulator off		0.01	1.0 2.0	μA μA
		$V_{EN} = 2.0V$, regulator on		0.15	1.0 2.0	μA μA
		$V_{EN} = 30V$, regulator on		0.5	2.5 5.0	μA μA

Note 1. Exceeding the absolute maximum rating may damage the device.

Note 2. The device is not guaranteed to function outside its operating rating.

Note 3: The absolute maximum positive supply voltage (60V) must be of limited duration ($\leq 100ms$) and duty cycle ($\leq 1\%$). The maximum continuous supply voltage is 30V.

Note 4: The maximum allowable power dissipation of any T_A (ambient temperature) is $P_{D(max)} = (T_{J(max)} - T_A) \div \theta_{JA}$. Exceeding the maximum allowable power dissipation will result in excessive die temperature, and the regulator will go into thermal shutdown. The θ_{JA} of the MIC5236-x.xBM (all versions) is $63^\circ C/W$, and the MIC5236-x.xBMM (all versions) is $80^\circ C/W$, mounted on a PC board (see "Thermal Characteristics" for further details).

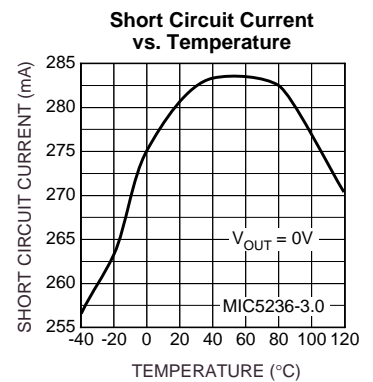
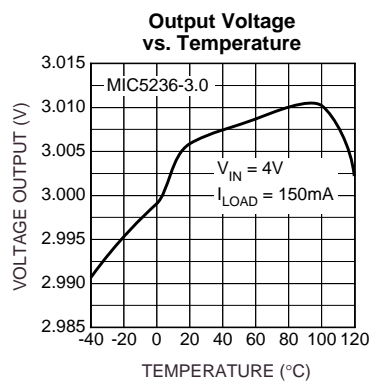
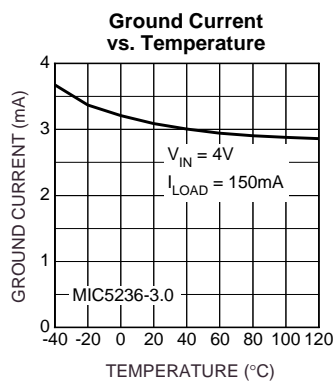
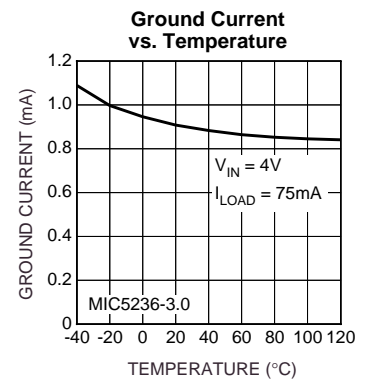
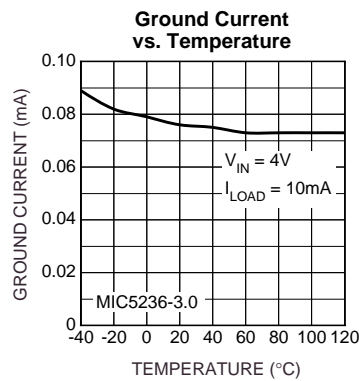
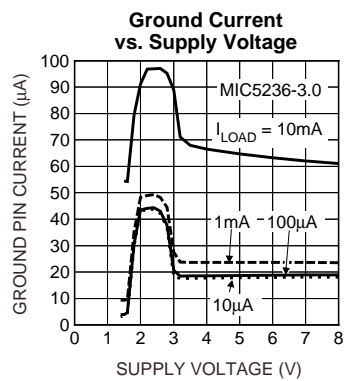
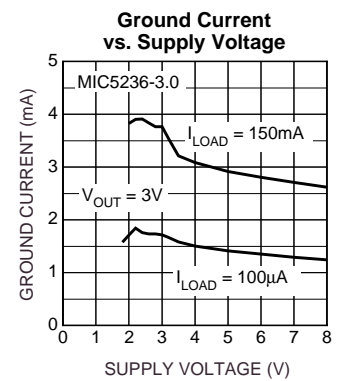
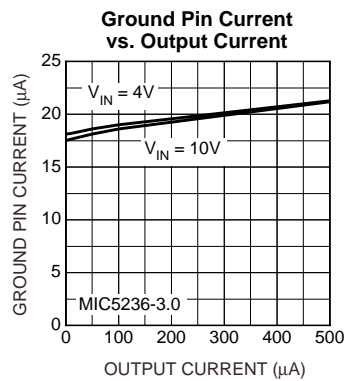
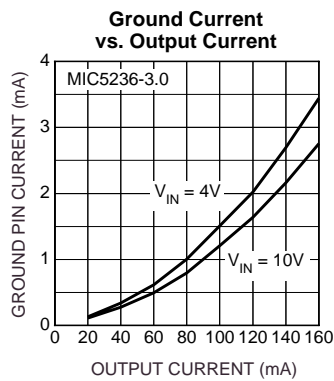
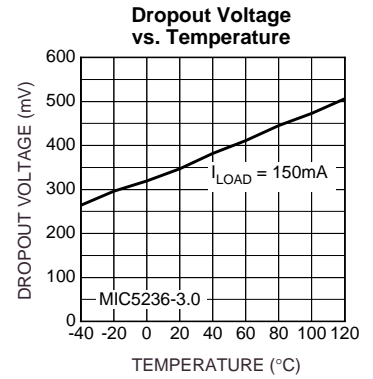
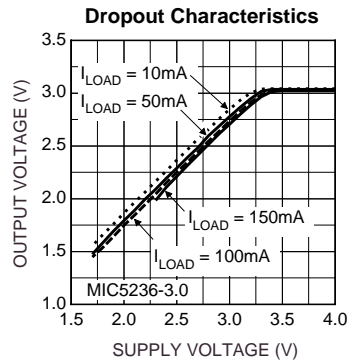
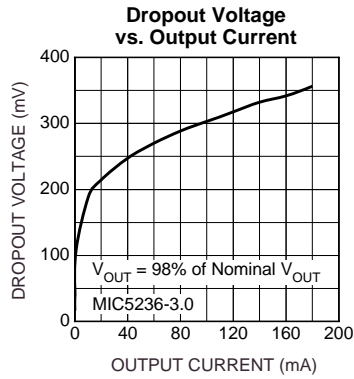
Note 5. Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5k in series with 100pF.

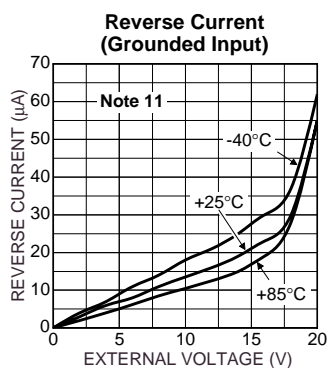
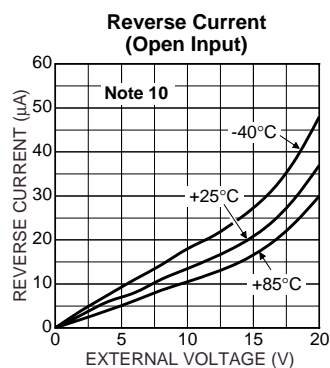
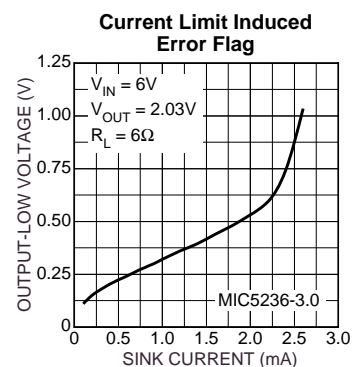
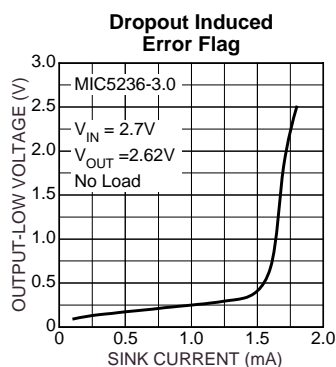
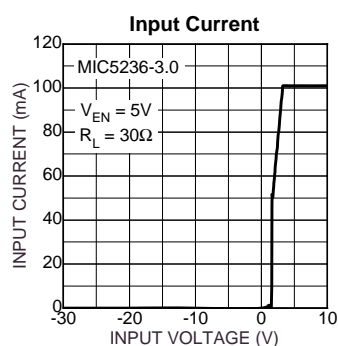
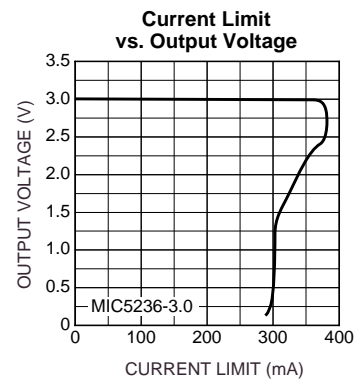
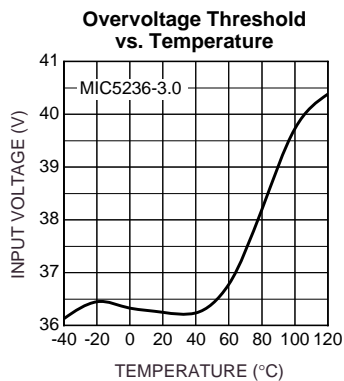
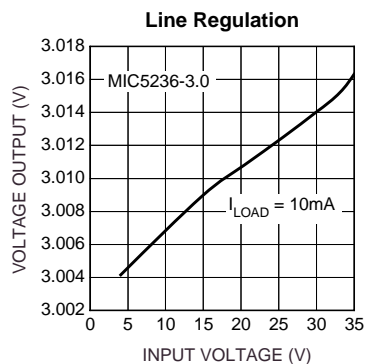
Note 6: Output voltage temperature coefficient is defined as the worst-case voltage change divided by the total temperature range.

Note 7: Regulation is measured at constant junction temperature using pulse testing with a low duty-cycle. Changes in output voltage due to heating effects are covered by the specification for thermal regulation.

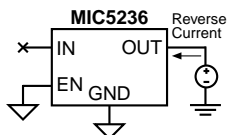
Note 8: Dropout voltage is defined as the input to output differential at which the output voltage drops 2% below its nominal value measured at 1.0V differential.

Typical Characteristics

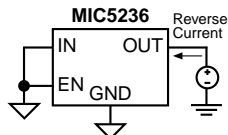




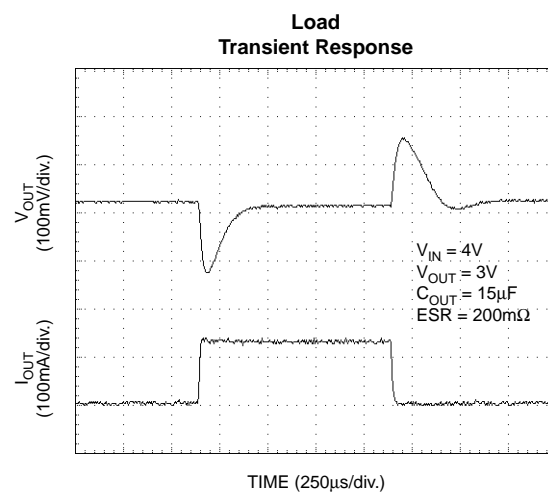
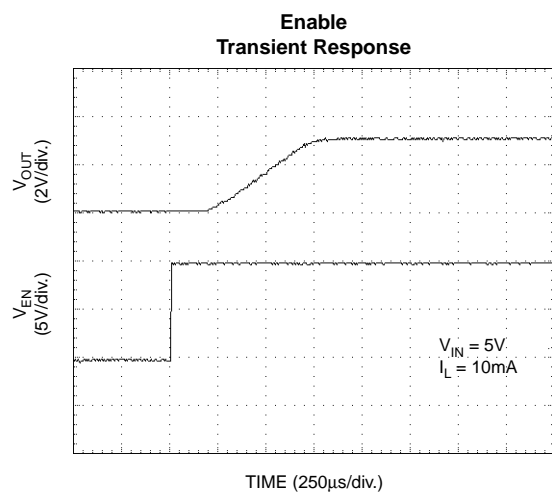
Note 10



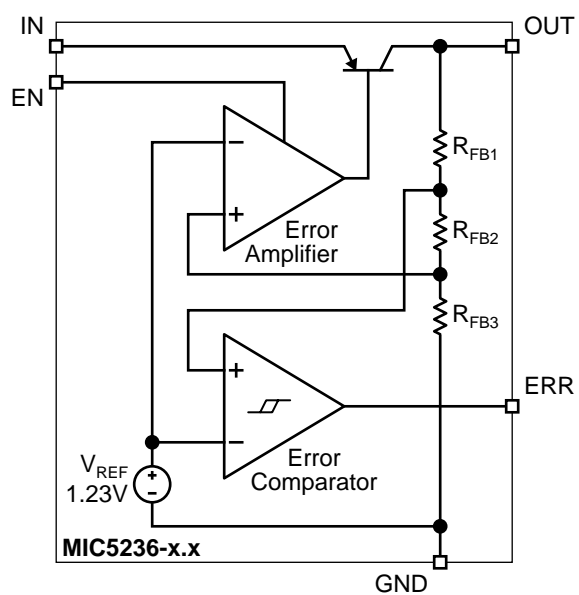
Note 11



Functional Characteristics



Functional Diagram



Application Information

The MIC5236 provides all of the advantages of the MIC2950: wide input voltage range, load dump (positive transients up to 60V), and reversed-battery protection, with the added advantages of reduced quiescent current and smaller package. Additionally, when disabled, quiescent current is reduced to 0.1 μ A.

Enable

A low on the enable pin disables the part, forcing the quiescent current to less than 0.1 μ A. Thermal shutdown and the error flag are not functional while the device is disabled. The maximum enable bias current is 2 μ A for a 2.0V input. An open collector pull-up resistor tied to the input voltage should be set low enough to maintain 2V on the enable input. Figure 1 shows an open collector output driving the enable pin through a 200k pull-up resistor tied to the input voltage.

In order to avoid output oscillations, slow transitions from low to high should be avoided.

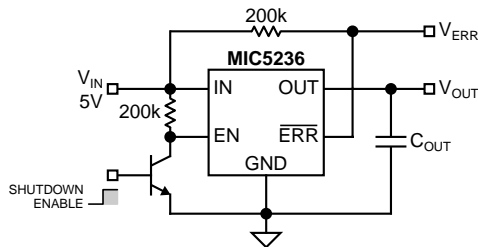


Figure 1. Remote Enable

Input Capacitor

An input capacitor may be required when the device is not near the source power supply or when supplied by a battery. Small, surface mount, ceramic capacitors can be used for bypassing. Larger values may be required if the source supply has high ripple.

Output Capacitor

The MIC5236 has been designed to minimize the effect of the output capacitor ESR on the closed loop stability. As a result, ceramic or film capacitors can be used at the output. Figure 2 displays a range of ESR values for a 10 μ F capacitor. Virtually any 10 μ F capacitor with an ESR less than 3.4 Ω is sufficient for stability over the entire input voltage range. Stability can also be maintained throughout the specified load and line conditions with 1 μ F film or ceramic capacitors.

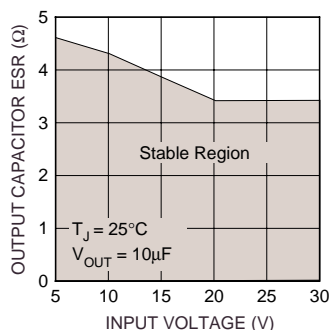


Figure 2. Output Capacitor ESR

Error Detection Comparator Output

The ERR pin is an open collector output which goes low when the output voltage drops 5% below its internally programmed level. It senses conditions such as excessive load (current limit), low input voltage, and over temperature conditions. Once the part is disabled via the enable input, the error flag output is not valid. Overvoltage conditions are not reflected in the error flag output. The error flag output is also not valid for input voltages less than 2.3V.

The error output has a low voltage of 400mV at a current of 200 μ A. In order to minimize the drain on the source used for the pull-up, a value of 200k to 1M Ω is suggested for the error flag pull-up. This will guarantee a maximum low voltage of 0.4V for a 30V pull-up potential. An unused error flag can be left unconnected.

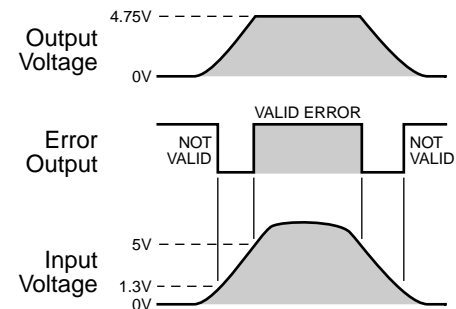


Figure 3. Error Output Timing

Reverse Current Protection

The MIC5236 is designed to limit the reverse current flow from output to input in the event that the MIC5236 output has been tied to the output of another power supply. See the graphs detailing the reverse current flow with the input grounded and open.

Thermal Shutdown

The MIC5236 has integrated thermal protection. This feature is only for protection purposes. The device should never be intentionally operated near this temperature as this may have detrimental effects on the life of the device. The thermal shutdown may become inactive while the enable input is transitioning a high to a low. When disabling the device via the enable pin, transition from a high to low quickly. This will insure that the output remains disabled in the event of a thermal shutdown.

Current Limit

Figure 4 displays a method for reducing the steady state short circuit current. The duration that the supply delivers current is set by the time required for the error flag output to discharge the 4.7 μ F capacitor tied to the enable pin. The off time is set by the 200K resistor as it recharges the 4.7 μ F capacitor, enabling the regulator. This circuit reduces the short circuit current from 280mA to 15mA while allowing for regulator restart once the short is removed.

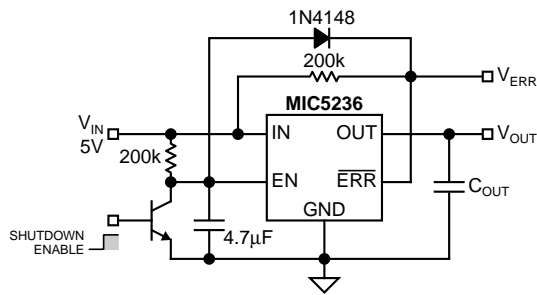


Figure 4. Remote Enable with Short-Circuit Current Foldback

Thermal Characteristics

The MIC5236 is a high input voltage device, intended to provide 150mA of continuous output current in two very small profile packages. The power SOP-8 and power MSOP-8 allow the device to dissipate about 50% more power than their standard equivalents.

Power SOP-8 Thermal Characteristics

One of the secrets of the MIC5236's performance is its power SO-8 package featuring half the thermal resistance of a standard SO-8 package. Lower thermal resistance means more output current or higher input voltage for a given package size.

Lower thermal resistance is achieved by joining the four ground leads with the die attach paddle to create a single-piece electrical and thermal conductor. This concept has been used by MOSFET manufacturers for years, proving very reliable and cost effective for the user.

Thermal resistance consists of two main elements, θ_{JC} (junction-to-case thermal resistance) and θ_{CA} (case-to-ambient thermal resistance). See Figure 5. θ_{JC} is the resistance from the die to the leads of the package. θ_{CA} is the resistance from the leads to the ambient air and it includes θ_{CS} (case-to-sink thermal resistance) and θ_{SA} (sink-to-ambient thermal resistance).

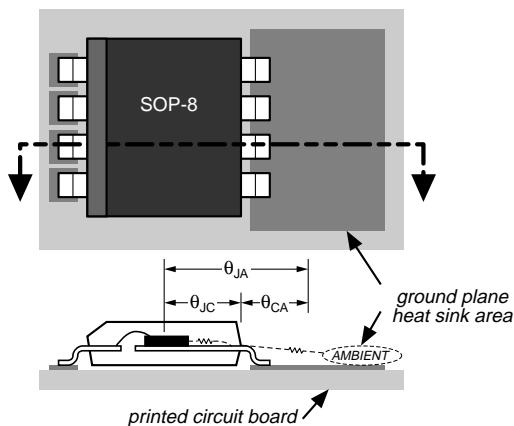


Figure 5. Thermal Resistance

Using the power SOP-8 reduces the θ_{JC} dramatically and allows the user to reduce θ_{CA} . The total thermal resistance, θ_{JA} (junction-to-ambient thermal resistance) is the limiting-factor in calculating the maximum power dissipation capability of the device. Typically, the power SOP-8 has a θ_{JC} of 20°C/W, this is significantly lower than the standard SOP-8 which is typically 75°C/W. θ_{CA} is reduced because pins 5

through 8 can now be soldered directly to a ground plane which significantly reduces the case-to-sink thermal resistance and sink to ambient thermal resistance.

Low-dropout linear regulators from Micrel are rated to a maximum junction temperature of 125°C. It is important not to exceed this maximum junction temperature during operation of the device. To prevent this maximum junction temperature from being exceeded, the appropriate ground plane heat sink must be used.

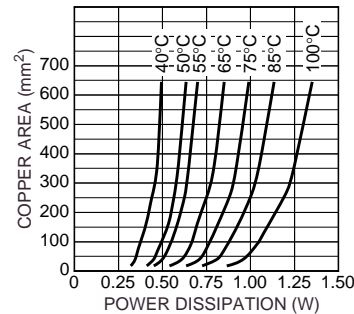


Figure 6. Copper Area vs. Power-SOP Power Dissipation (ΔT_{JA})

Figure 6 shows copper area versus power dissipation with each trace corresponding to a different temperature rise above ambient.

From these curves, the minimum area of copper necessary for the part to operate safely can be determined. The maximum allowable temperature rise must be calculated to determine operation along which curve.

$$\Delta T = T_{J(max)} - T_{A(max)}$$

$$T_{J(max)} = 125^{\circ}\text{C}$$

$$T_{A(max)} = \text{maximum ambient operating temperature}$$

For example, the maximum ambient temperature is 50°C, the ΔT is determined as follows:

$$\Delta T = 125^{\circ}\text{C} - 50^{\circ}\text{C}$$

$$\Delta T = 75^{\circ}\text{C}$$

Using Figure 6, the minimum amount of required copper can be determined based on the required power dissipation. Power dissipation in a linear regulator is calculated as follows:

$$P_D = (V_{IN} - V_{OUT}) I_{OUT} + V_{IN} \cdot I_{GND}$$

If we use a 3V output device and a 28V input at moderate output current of 25mA, then our power dissipation is as follows:

$$P_D = (28\text{V} - 3\text{V}) \times 25\text{mA} + 28\text{V} \times 250\mu\text{A}$$

$$P_D = 625\text{mW} + 7\text{mW}$$

$$P_D = 632\text{mW}$$

From Figure 6, the minimum amount of copper required to operate this application at a ΔT of 75°C is 25mm².

Quick Method

Determine the power dissipation requirements for the design along with the maximum ambient temperature at which the device will be operated. Refer to Figure 7, which shows safe operating curves for three different ambient temperatures: 25°C, 50°C and 85°C. From these curves, the minimum

amount of copper can be determined by knowing the maximum power dissipation required. If the maximum ambient temperature is 50°C and the power dissipation is as above, 632mW, the curve in Figure 7 shows that the required area of copper is 25mm².

The θ_{JA} of this package is ideally 63°C/W, but it will vary depending upon the availability of copper ground plane to which it is attached.

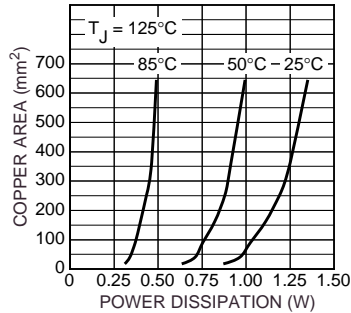


Figure 7. Copper Area vs. Power-SOP Power Dissipation (T_A)

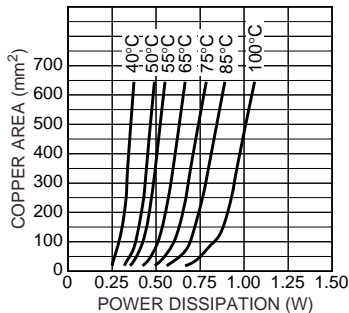


Figure 8. Copper Area vs. Power-MSOP Power Dissipation (ΔT_{JA})

The same method of determining the heat sink area used for the power-SOP-8 can be applied directly to the power-MSOP-8. The same two curves showing power dissipation versus copper area are reproduced for the power-MSOP-8 and they can be applied identically, see Figures 8 and 9.

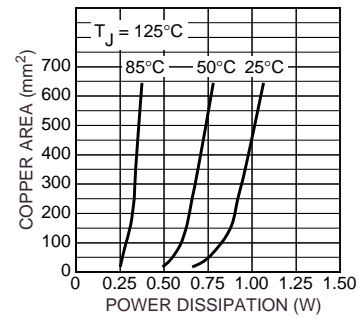


Figure 9. Copper Area vs. Power-MSOP Power Dissipation (T_A)

Power MSOP-8 Thermal Characteristics

The power-MSOP-8 package follows the same idea as the power-SO-8 package, using four ground leads with the die attach paddle to create a single-piece electrical and thermal conductor, reducing thermal resistance and increasing power dissipation capability.

Quick Method

Determine the power dissipation requirements for the design along with the maximum ambient temperature at which the device will be operated. Refer to Figure 9, which shows safe operating curves for three different ambient temperatures, 25°C, 50°C, and 85°C. From these curves, the minimum amount of copper can be determined by knowing the maximum power dissipation required. If the maximum ambient temperature is 50°C, and the power dissipation is 639mW, the curve in Figure 9 shows that the required area of copper is 110mm², when using the power MSOP-8.

Adjustable Regulator Application

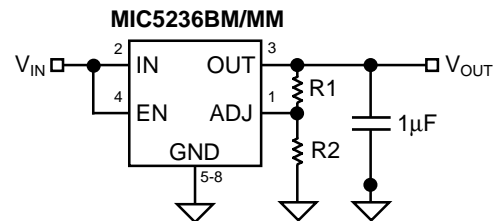


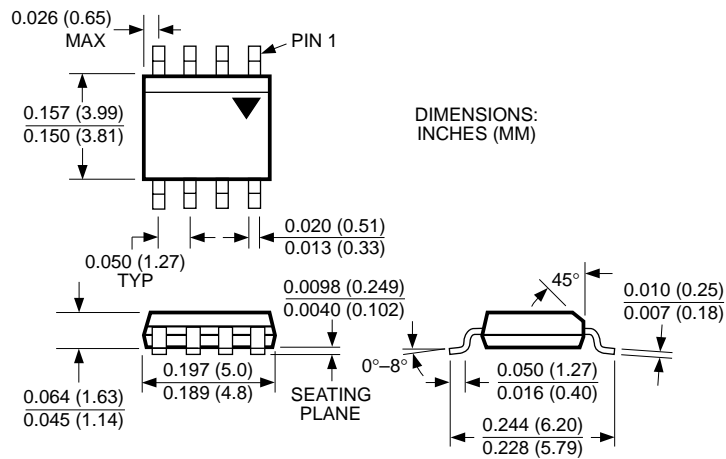
Figure 10. Adjustable Voltage Application

The MIC5236BM/MM can be adjusted from 1.24V to 20V by using two external resistors (Figure 10). The resistors set the output voltage based on the following equation:

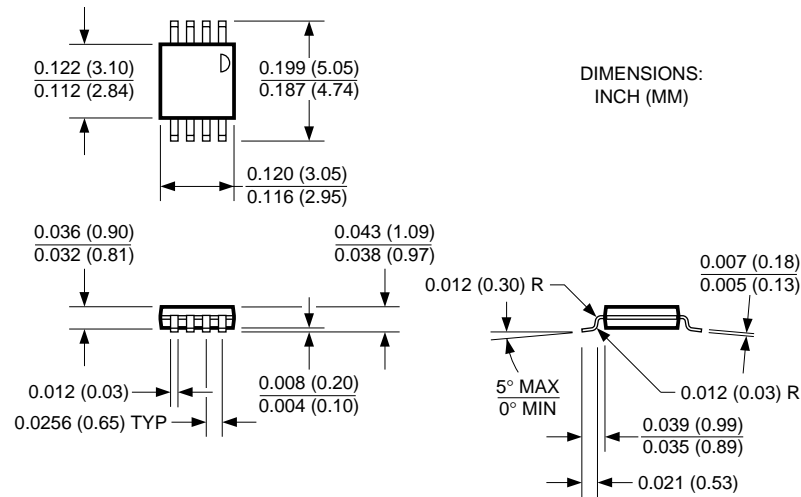
$$V_{OUT} = V_{REF} \left(1 + \frac{R1}{R2} \right)$$

Where $V_{REF} = 1.23V$.

Package Information



8-Lead SOIC (M)



8-Lead MSOP (MM)

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